

## **CMOS IC Application Note**

# S-19942/19943 Series NOISE COUNTERMEASURES AND CISPR25 MEASUREMENT RESULTS

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This application note is a reference describing recommended noise suppression parts and board layouts that help reduce conductive noise and emission noise for the S-19942/19943 Series. It also summarizes CISPR25 compliant measurement results.

Refer to the datasheet for details and specs.

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#### 1. Need for Noise Suppression Parts

**Figure 1** shows the results of a voltage method measurement of a circuit to which no noise suppression parts have been applied. Although the measurement results significantly exceeded the CISPR25 class 5 specifications, this is not limited to this IC as general switching regulator ICs exhibit the similar tendency. For this reason, noise suppression parts must be applied to allow the IC to meet the stringent CISPR25 class 5 requirements. **Figure 1** to **Figure 3** show the necessity of noise suppression parts using our previous product as an example.

Our previous product (2.2 MHz), V<sub>IN</sub> = 13 V, V<sub>OUT</sub> = 5 V, I<sub>OUT</sub> = 600 mA, without noise suppression parts

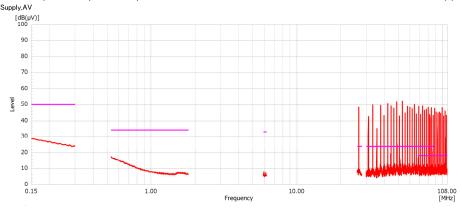


Figure 1

**Figure 2** is a circuit diagram to which the recommended noise suppression parts L2, C<sub>F2</sub>, and C<sub>F4</sub> listed in **Table 1** have been added. **Figure 3** shows the results of a voltage method measurement carried out on this circuit. The addition of noise suppression parts reduces noise by as much as 20 MHz or more allowing the IC to meet CISPR25 class 5 requirements. Also, this IC optimally adjusts the SW pin slew rate in the IC so that the minimum of noise suppression parts will obtain a substantial margin.

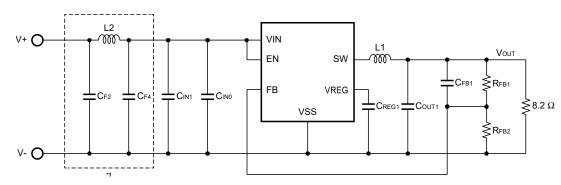


Table 1		
Symbol	Value	
C <sub>F2</sub>	4.7 μF	
L2	4.7 μΗ	
C <sub>F4</sub>	4.7 μF	

#### **\*1.** Noise suppression parts

Figure 2

Our previous product (2.2 MHz), V<sub>IN</sub> = 13 V, V<sub>OUT</sub> = 5 V, I<sub>OUT</sub> = 600 mA, with noise suppression parts

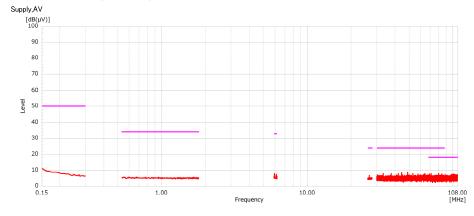


Figure 3

This IC has a built-in spread spectrum clock generation circuit capable of reducing conductive noise and emission noise. Refer to the datasheet of the S-19942A/19943B/19943B Series for details.

#### 2. Board Layout

#### 2. 1 Placement of input capacitors (C<sub>IN0</sub>, C<sub>IN1</sub>) and VIN and VSS layouts

The placement of input capacitors and VIN and VSS wirings are crucial. The input capacitors are of paramount importance and should be placed as close to the IC as possible and on the same surface layer.

 $C_{\text{IN1}}$  is an essential capacitor to ensure stable IC operation and to suppress noise.  $C_{\text{IN0}}$  is a capacitor of approximately 0.1  $\mu\text{F}$  which is connected in parallel to  $C_{\text{IN1}}$ . Add this capacitor as required to suppress mainly 10 MHz or larger noise. **Figure 4** shows an example of placing input capacitors in the immediate vicinity of the IC. The bold line indicates the current path the instant the high side power MOS FET in the IC turns on. First, current flows from the input capacitors to the VIN pin. Next, the current goes through the high side power MOS FET which is ON and the parasitic capacitance of low side power MOS FET which is OFF, both of which are inside the IC, in turn. Finally, the current returns to the input capacitors from the VSS pin. In this example, the reduction of impedance in the current path minimizes the noise generated in VIN and VSS wirings.

By contrast, **Figure 5** shows an example of input capacitors placed away from the IC and where the VSS pin is not connected to the VSS directly below the IC. At this time, there is a parasitic capacitance ( $L_{p1}$ ) between the input capacitors and the VIN pin. In addition, as the current path between the input capacitors and the VSS pin is significantly extended, the parasitic capacitances ( $L_{p2} + L_{p3} + L_{p4}$ ) extremely become large. In this example, impedance in the current path is increased and the noise generated in VIN and VSS wirings become larger.

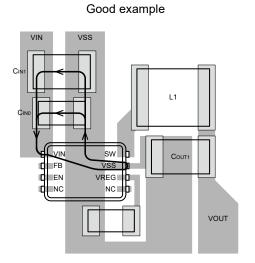


Figure 4

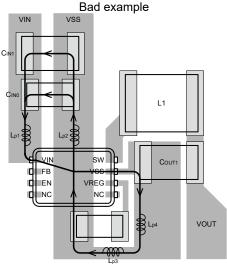


Figure 5

#### 2. 2 Placement and layout of Inductor (L1)

The wiring area of the SW pin should be minimized. Reduce the wiring length within the allowable current capacity range. The rectangular wave voltage output from the SW pin contains high-frequency components, which could cause SW wiring to act as an antenna and thereby increase emission noise. In addition, as the high-frequency components in the rectangular wave voltage previously mentioned are transmitted from the SW pin to VOUT via the parasitic capacitance, SW wiring must be kept apart from VOUT wiring. Select a closed magnetic path inductor with low emission noise.

**Figure 6** shows an example where the wiring area between the SW pin and L1 is reduced and the distance between SW wiring and VOUT wiring is extended.

By contrast, **Figure 7** shows an example where the SW wiring area is excessively large, the distance between SW wiring and VOUT wiring is short, and the parasitic capacitance  $(C_p)$  is large.

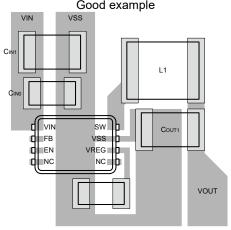


Figure 6

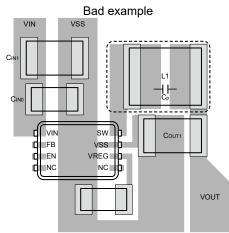


Figure 7

#### 2. 3 Placement and layout of output capacitor (Cout1)

Place C<sub>OUT1</sub> close to the IC.

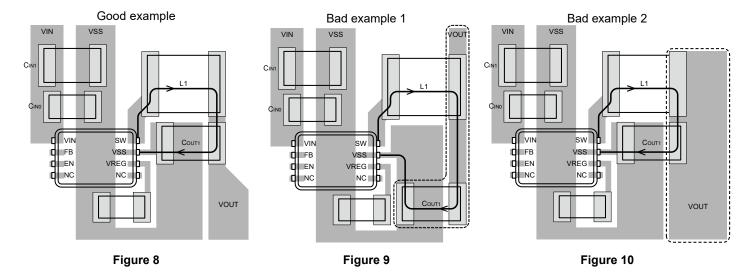
If the area of the current path indicated by the bold line (SW pin  $\rightarrow$  L1  $\rightarrow$  C<sub>OUT1</sub>  $\rightarrow$  VSS pin) is reduced, the emission noise to be generated will be minimized.

Be sure to pull out VOUT wiring after routing through the  $C_{OUT1}$  land. If this is not the case, smoothing due to L1 and  $C_{OUT1}$  is weakened causing high-frequency components in the rectangular wave voltage of SW pin to be conducted to VOUT. Similarly, when increasing the line width, pull out a wiring after routing through  $C_{OUT1}$  land.

**Figure 8** is an example where the current path area is narrowed. VOUT wiring is pulled out after routing through the C<sub>OUT1</sub> land.

By contrast, **Figure 9** is an example where the distance between the IC and  $C_{OUT1}$  is long, and the current path area is large. VOUT wiring is pulled out from L1 land without routing through  $C_{OUT1}$  land.

Figure 10 is an example where VOUT line width is excessively increased before routing through Couting through



#### 2. 4 Recommended board layout

Figure 11 shows the recommended board layout that reflects the explanations previously mentioned.

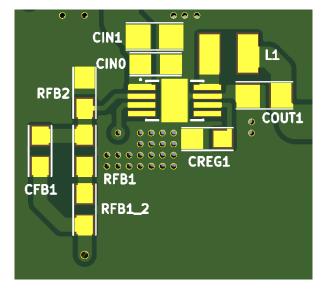


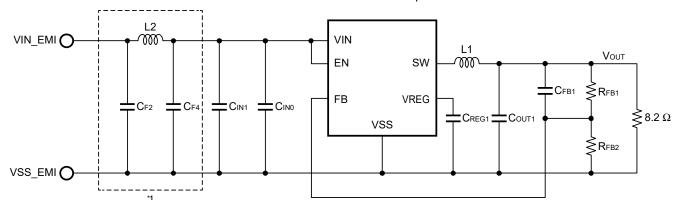
Figure 11

#### 3. Measurement Conditions and Measurement Results

#### 3. 1 Measurement condition

Figure 12 shows a circuit diagram of S-19943A Evaluation Board. The parts listed in Table 2 are mounted.

S-19943A,  $V_{IN}$  = 13 V,  $V_{OUT}$  = 5 V,  $I_{OUT}$  = 600 mA, Compliant with CISPR25 Edition 4.0 2016-10



\*1. Noise suppression parts

Figure 12 Circuit Diagram of S-19943A Evaluation Board

Symbol Value Part Number Manufacturer C<sub>F2</sub> 4.7 μF CGA4J1X7R1H475K125AC **TDK Corporation** L2 TFM252012ALVA4R7MTAA  $4.7 \, \mu H$ **TDK Corporation** C<sub>F4</sub>  $4.7~\mu F$ CGA4J1X7R1H475K125AC **TDK Corporation** CGA4J1X7R1H475K125AC C<sub>IN1</sub> 4.7 μF **TDK Corporation** CINO  $0.1 \, \mu F$ CGA3E2X7R1H104K **TDK Corporation** TFM252012ALVA4R7MTAA L1  $4.7 \mu H$ **TDK Corporation** CGA3E1X7R1C105K080AC C<sub>REG1</sub> 1.0 μF **TDK Corporation** C<sub>OUT1</sub> CGA4J3X7S1A106K125AB 10 μF **TDK Corporation** Сғв 33 pF R<sub>FB1</sub> 84 kΩ  $R_{FB2}$ 16 k $\Omega$ 

Table 2

Figure 13 shows the actual board.

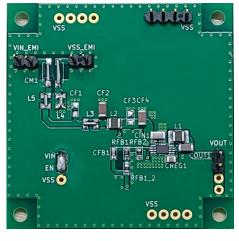
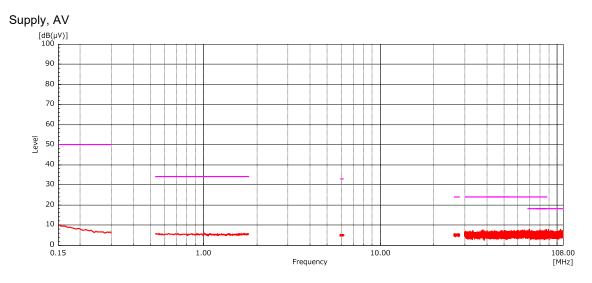


Figure 13

#### 3. 2 Measurement results of voltage method

Figure 14 shows measurement results of voltage method.

S-19943A Supply: VIN\_EMI, Return: VSS\_EMI, AV: average





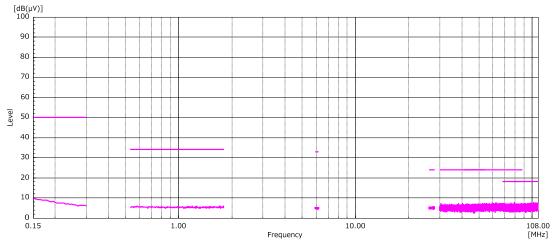


Figure 14

#### 3. 3 Measurement results of ALSE method

Figure 15 and Figure 16 show measurement results of ALSE method.

S-19943A V: vertical, H: horizontal, AV: average

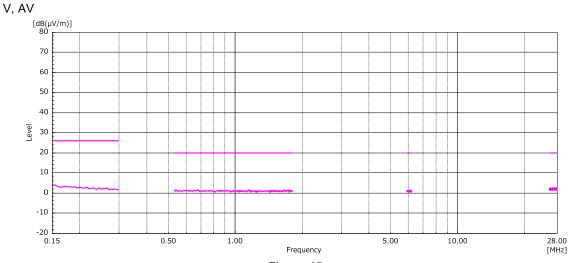
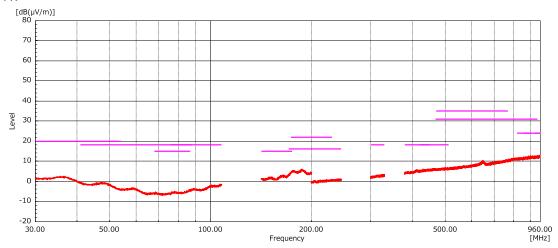


Figure 15







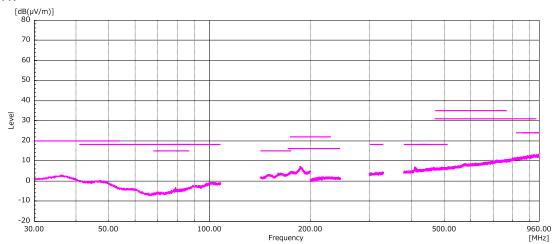


Figure 16

#### 4. Precautions

- The usages described in this application note are typical examples using ICs of ABLIC Inc. Perform thorough evaluation before use.
- Do not apply an electrostatic discharge to this IC that exceeds the performance ratings of the built-in electrostatic protection circuit.
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#### 5. Related Sources

Refer to the following datasheet for details of the S-19942/19943 Series.

#### S-19942A/19942B/19943A/19943B Series Datasheet

The information described in this application note and the datasheet is subject to change without notice. Contact our sales representatives for details.

Regarding the newest version of the datasheet, select product category and product name on our website, and download the PDF file.

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