



CMOS IC Application Note

S-19560B Series Application Examples and Component Selection Methods

Rev.1.0_00

© ABLIC Inc., 2025

This application note is a method of external parts selection and recommended board layouts for the S-19560B Series. Refer to the datasheet for details and specs.

Contents

1.	Ch1 Output Current Calculation Method	3
1. 1	Step-down DC-DC converter input current calculation method	3
1. 2	LDO regulator input current calculation method	4
1. 3	S-19560B Series VOUT1 pin input current calculation.....	5
2.	Relationship between Loss and Heat Generation	6
2. 1	Ch1, Ch2 (DC-DC converter) loss	6
2. 2	Ch3 (LDO regulator) loss	6
2. 3	Heat generation amount calculation	6
3.	Typical Usage Examples	7
3. 1	Example use as a power supply for a camera module	7
3. 2	Example where Ch2 output is unused.....	8
3. 3	Example where Ch3 output is unused.....	8
4.	Inductor Selection.....	9
4. 1	Inductor peak current calculation	9
4. 2	VOUT ripple voltage calculation	9
4. 3	Step-down DC-DC converter duty calculation	9
4. 4	Specific product examples with an inductance value of 2.2 μ H.....	9
4. 5	Reference data (inductance value 2.2 μ H)	10
5.	Board Layout Precautions	14
5. 1	Application circuit diagram	14
5. 2	PCB board layout	14
5. 3	Board layout precautions.....	15
6.	Precautions	16
7.	Related Sources	16

1. Ch1 Output Current Calculation Method

The S-19560B Series power supply configuration is shown in **Figure 1**.

The VOUT1 pin serves as both the Ch1 feedback pin (FB) and the Ch2 and Ch3 power supply. As such, note that it is not possible to supply the entire Ch1 output current (I_{OUT1}) to the load.

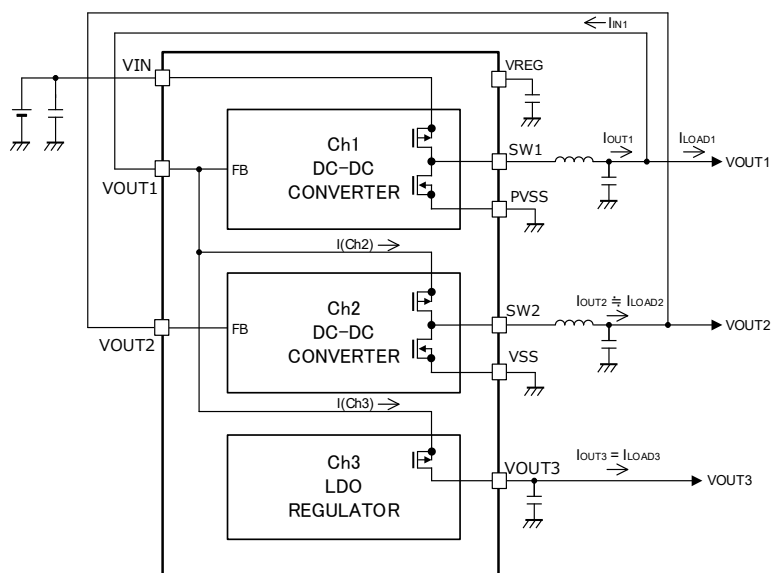


Figure 1 Power Supply Configuration Diagram

1.1 Step-down DC-DC converter input current calculation method

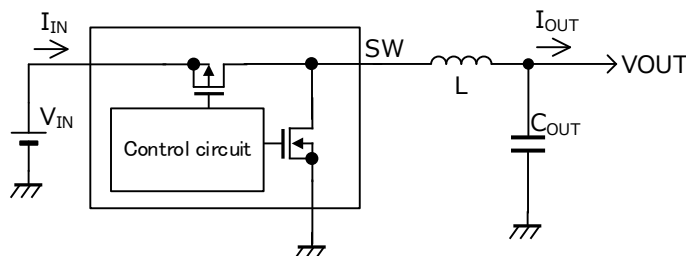


Figure 2 Standard Step-down DC-DC Converter

Each of the variables in **Figure 2** are defined as follows.

V_{IN} :	Input voltage
I_{IN} :	Input current
V_{OUT} :	Output voltage
I_{OUT} :	Output current
η :	Conversion efficiency

Input power \times efficiency = output power, so

$$V_{IN} \times I_{IN} \times \eta^* = V_{OUT} \times I_{OUT}$$

The DC-DC converter input current is

$$I_{IN} = \frac{V_{OUT} \times I_{OUT}}{V_{IN} \times \eta} \dots\dots\dots (1)$$

*1. The control circuit self-power consumption is reflected in η .

1.2 LDO regulator input current calculation method

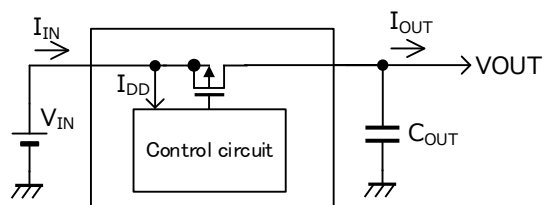


Figure 3 Standard LDO Regulator

Each of the variables in **Figure 3** are defined as follows.

V_{IN} :	Input voltage
I_{IN} :	Input current
I_{DD} :	Self-consumption current
V_{OUT} :	Output voltage
I_{OUT} :	Output current

The LDO regulator input current is

$$I_{IN} = I_{OUT} + I_{DD}$$

Here, $I_{OUT} > I_{DD}$, so

$$I_{IN} \approx I_{OUT} \dots\dots\dots (2)$$

1.3 S-19560B Series VOUT1 pin input current calculation

The power supply for the S-19560B Series Ch2 (DC-DC converter) and Ch3 (LDO regulator) are supplied from the VOUT1 pin.

In order to determine the allowable current level (I_{LOAD1}) that can be supplied from the VOUT1 output voltage to the load, it is necessary to calculate the input currents of Ch2 and Ch3.

In addition, because the VOUT1 pin also functions as the Ch1 feedback pin, a separate power supply cannot be connected.

Each of the variables in the following formulas are defined as follows. Refer to **Figure 1** for the relationships between I_{OUT1} , I_{OUT2} , I_{OUT3} , I_{LOAD1} , I_{LOAD2} , and I_{LOAD3} .

V_{OUT1} :	VOUT1 output voltage
V_{OUT2} :	VOUT2 output voltage
I_{OUT1} :	Ch1 output current
I_{OUT2} :	Ch2 output current
I_{OUT3} :	Ch3 output current
I_{LOAD1} :	VOUT1 load current
I_{LOAD2} :	VOUT2 load current
I_{LOAD3} :	VOUT3 load current
η_2 :	Ch2 conversion efficiency

As such, if the Ch2 input current ($I_{(Ch2)}$) and Ch3 input current ($I_{(Ch3)}$) are calculated using formulas (1) and (2), then the VOUT2 pin input current will be extremely small, so

$$I_{OUT2} \approx I_{LOAD2}$$

therefore, the following formula is derived from formula (1).

$$I_{(Ch2)} = \frac{V_{OUT2} \times I_{LOAD2}}{V_{OUT1} \times \eta_2}$$

From **Figure 1**

$$I_{OUT3} = I_{LOAD3}$$

therefore, the following formula is derived from formula (2).

$$I_{(Ch3)} \approx I_{LOAD3}$$

The VOUT1 pin input current (I_{IN1}) will be as follows.

$$\begin{aligned} I_{IN1} &= I_{(Ch2)} + I_{(Ch3)} \\ &= \frac{V_{OUT2} \times I_{LOAD2}}{V_{OUT1} \times \eta_2} + I_{LOAD3} \end{aligned}$$

The maximum Ch1 output current (I_{OUT1}) is 600 mA. The maximum current ($I_{LOAD1(max)}$) which Ch1 can directly supply to the load is this value minus the current (I_{IN1}) supplied to Ch2 and Ch3.

$$I_{LOAD1(max)} = 600 \text{ mA} - I_{IN1}$$

2. Relationship between Loss and Heat Generation

Loss generated in the S-19560B Series becomes heat, increasing the temperature of the junction. Therefore, the estimation of loss is extremely important.

Each of the variables in the following formulas are defined as follows.

V _{OUT1} :	VOUT1 output voltage
V _{OUT2} :	VOUT2 output voltage
I _{OUT1} :	Ch1 output current
I _{OUT2} :	Ch2 output current
I _{OUT3} :	Ch3 output current
η ₁ :	Ch1 conversion efficiency
η ₂ :	Ch2 conversion efficiency

2.1 Ch1, Ch2 (DC-DC converter) loss

The main sources of loss in a DC-DC converter are losses generated inside the IC and loss generated by external components. The total loss value can be estimated from the efficiency characteristics.

The loss for Ch1 and Ch2 (P_{LOSS1}, P_{LOSS2}) can be calculated from the following formulas.

$$P_{LOSS1} = V_{OUT1} \times I_{OUT1} \times \left(\frac{1 - \eta_1}{\eta_1} \right)$$

$$P_{LOSS2} = V_{OUT2} \times I_{OUT2} \times \left(\frac{1 - \eta_2}{\eta_2} \right)$$

Estimate loss within the output current range that will actually be used.

2.2 Ch3 (LDO regulator) loss

The main source of loss in an LDO regulator can be found by multiplying the input-output voltage difference by the output current.

The loss for Ch3 (P_{LOSS3}) can be calculated from the following formula.

$$P_{LOSS3} = (V_{OUT1} - V_{OUT3}) \times I_{OUT3}$$

For the S-19560B Series, reducing the V_{OUT1} output voltage will decrease the Ch3 (LDO regulator) loss. Determine the setting value for V_{OUT1} output voltage while taking the following conditions into consideration.

- Possible setting range: 3.3 V to 5.5 V
- At least 0.7 V higher than the V_{OUT2} setting voltage
- At least 0.3 V higher than the V_{OUT3} setting voltage
- If connecting an external LDO regulator, ensure that the input-output voltage difference required for the external LDO regulator is secured.

2.3 Heat generation amount calculation

Overall loss in the S-19560B Series can be calculated from the following formula.

$$P_{LOSS} = P_{LOSS1} + P_{LOSS2} + P_{LOSS3}$$

The junction temperature (T_J) caused by the generated loss P_{LOSS} can be calculated from the following formula. The thermal resistance value (θ_{JA}) is shown in **Table 1**.

$$T_J = P_{LOSS} \times \theta_{JA} + T_a$$

Table 1

Item	Symbol	Condition	Min.	Typ.	Max.	単位	
Junction-to-ambient thermal resistance*1	θ_{JA}	HSNT-8(2030)	Board A	-	181	-	°C/W
			Board B	-	135	-	°C/W
			Board C	-	40	-	°C/W
			Board D	-	42	-	°C/W
			Board E	-	32	-	°C/W

*1. Test environment: compliance with JEDEC STANDARD JESD51-2A

Remark Refer to "■ Power Dissipation" and "Test Board" of S-19560B Series datasheet for details.

3. Typical Usage Examples

3.1 Example use as a power supply for a camera module

The input voltage of 4 V to 16 V from VIN is first stepped down to approximately 3.3 V to 3.6 V in Ch1, then output to VOUT1.

The VOUT1 output voltage is reduced to low noise in the S-19255 series (LDO regulator)*1 and used as an analog power supply for the image sensor. For the Core power supply, which requires the most current, the VOUT2 output voltage which is the output from Ch2 is used to support the high current requirement. The VOUT3 output voltage which is the Ch3 output is used as the power supply for the I/O and serializer.

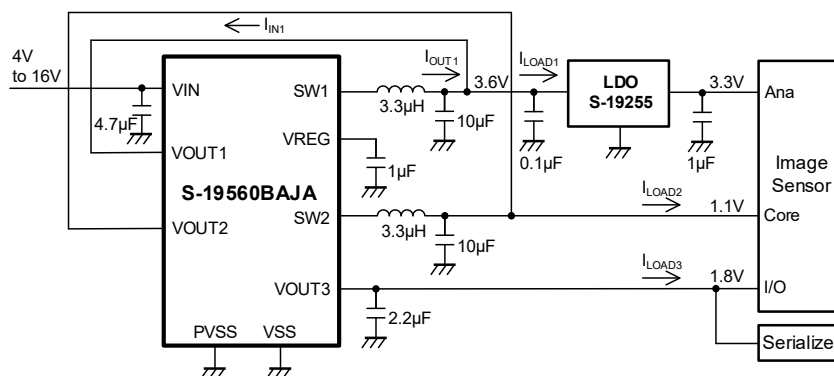


Figure 4

【Design example】

$$V_{IN} = 6.0 \text{ V}$$

S-19560BAJA

$$V_{OUT1} = 4.0 \text{ V}$$

$$V_{OUT2} = 1.1 \text{ V}$$

$$V_{OUT3} = 1.8 \text{ V}$$

$$I_{LOAD1} = 200 \text{ mA}$$

$$I_{LOAD2} = 400 \text{ mA}$$

$$I_{LOAD3} = 200 \text{ mA}$$

$$\eta_1 = 87\%^2$$

$$\eta_2 = 84\%^2$$

$$\theta_{JA} = 32^\circ\text{C/W}$$

$$I_{IN1} = \frac{V_{OUT2} \times I_{LOAD2}}{V_{OUT1} \times \eta_2} + I_{LOAD3}$$

$$= 331 \text{ mA}$$

$$I_{OUT1} = I_{IN1} + I_{LOAD1}$$

$$= 531 \text{ mA}$$

$$P_{LOSS} = P_{LOSS1} + P_{LOSS2} + P_{LOSS3}$$

$$= V_{OUT1} \times I_{OUT1} \times \left(\frac{1 - \eta_1}{\eta_1} \right) + V_{OUT2} \times I_{OUT2} \times \left(\frac{1 - \eta_2}{\eta_2} \right) + (V_{OUT1} - V_{OUT3}) \times I_{OUT3}$$

$$= 317 \text{ mW} + 83 \text{ mW} + 440 \text{ mW}$$

$$= 841 \text{ mW}$$

Increasing junction temperature

$$\Delta T_j = 0.841 \times 32 = 26.9^\circ\text{C}$$

*1. Refer to S-19255 Series datasheet for details.

*2. Refer to "■ Reference Data" of S-19560B Series datasheet for details.

Caution The above are theoretical calculation values. Perform thorough evaluation and verify characteristics in the actual application.

3.2 Example where Ch2 output is unused

VOUT2 is the feedback input pin for Ch2. The SW2 pin will be either "L" or "H" level depending on the relationship between the VOUT2 setting voltage value and the actual VOUT2 pin voltage. Therefore, if a voltage close to the setting value is applied to the VOUT2 pin, the SW2 pin output may repeatedly switch between "L" and "H". So, connecting VOUT2 to VOUT1, which has a sufficiently higher voltage, the SW2 pin voltage is fixed at "L" and becomes stable.

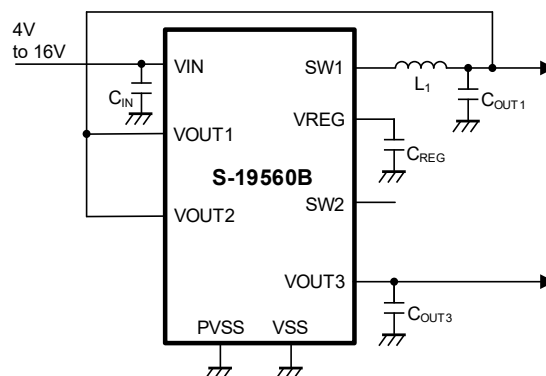


Figure 5

3.3 Example where Ch3 output is unused

If an output capacitance is not connected to the VOUT3 pin, the VOUT3 output voltage may become unstable, becoming an unintended source of noise. Even if Ch3 is not used, connect COUT3. If COUT3 must be left unconnected, connect VOUT3 to VOUT1. This will prevent the VOUT3 output voltage from becoming a noise source.

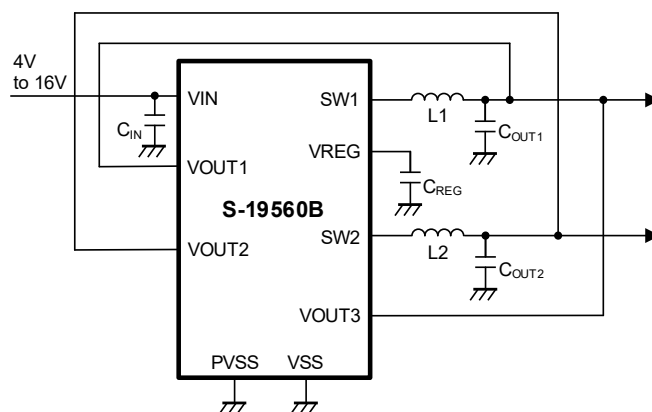


Figure 6

4. Inductor Selection

The recommended inductance value for the S-19560B Series is 3.3 μH . If a 2.2 μH inductor can be used, the mounting area can be reduced. In general, reducing the inductance value will increase the ripple components which flow through the coil, resulting in a higher peak current (I_{PK}) even at the same I_{OUT} . As such, the current limit value may be exceeded in some conditions.

In addition, the ripple voltage generated in the output will also increase.

Furthermore, there will also be an increased likelihood of subharmonic oscillation, which is a characteristic phenomenon of DC-DC converters in current mode. It is recommended to use in conditions where Duty < 50% in order to reduce the risk of subharmonic oscillation.

4.1 Inductor peak current calculation

The peak current (I_{PK}) which flows to a standard step-down DC-DC converter inductor shown in **Figure 2** can be calculated from the following formula.

f_{OSC} represents the switching frequency.

$$\Delta I_L = \frac{V_{\text{OUT}} \times (V_{\text{IN}} - V_{\text{OUT}})}{f_{\text{OSC}} \times L \times V_{\text{IN}}}$$

$$I_{\text{PK}} = I_{\text{OUT}} + \frac{\Delta I_L}{2}$$

4.2 VOUT ripple voltage calculation

$$\Delta V_{\text{OUT}} = \frac{\Delta I_L}{8 \times f_{\text{OSC}} \times C_{\text{OUT}}}$$

4.3 Step-down DC-DC converter duty calculation

$$\text{Duty} = \frac{V_{\text{OUT}}}{V_{\text{IN}}}$$

4.4 Specific product examples with an inductance value of 2.2 μH

Table 2

Manufacturer	Part Number	Withstanding Voltage	Temperature Range	Dimensions (L × W × H)
TDK Corporation	TFM201210ALMA2R2MTAA	20 V	-55°C to 150°C	2.0 mm × 1.25 mm × 1.0 mm
TDK Corporation	TFM201610ALMA2R2MTAA	20 V	-55°C to 150°C	2.0 mm × 1.6 mm × 1.0 mm
Murata Manufacturing Co., Ltd.	DfE2MCAH2R2MJ0	40 V	-40°C to 150°C	2.0 mm × 1.6 mm × 1.2 mm
TAIYO YUDEN CO., LTD.	LCCNF2012KKT2R2MAD	20 V	-55°C to 150°C	2.0 mm × 1.25 mm × 1.0 mm
TAIYO YUDEN CO., LTD.	LCENA2016MKT2R2M0TK	20 V	-40°C to 150°C	2.0 mm × 1.6 mm × 1.2 mm

4.5 Reference data (Inductance value 2.2 μH)

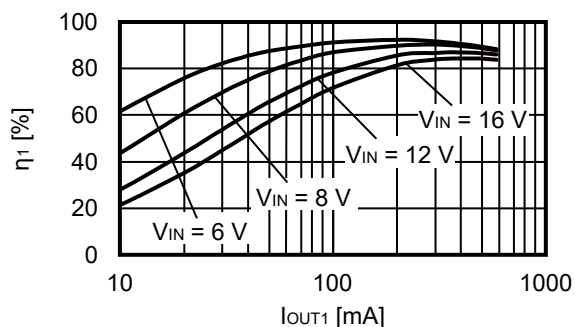
The external parts shown in **Table 3** are used in the reference data. Refer to "■ Reference Data" on the S-19560B Series datasheet for reference data for the inductance value of 3.3 μH .

Table 3

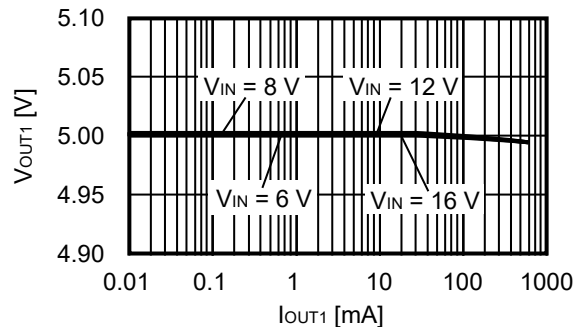
Symbol	Part Number	Constant	Withstanding Voltage	Dimensions (L × W × H)	Manufacturer
C _{IN}	CGA4J1X7R1E475K125AC	4.7 μF	25 V	2.0 mm × 1.25 mm × 1.25 mm	TDK Corporation
C _{OUT1} , C _{OUT2}	CGA4J3X7S1A106K125AB	10 μF	10 V	2.0 mm × 1.25 mm × 1.25 mm	TDK Corporation
C _{OUT3}	CGA3E1X7R0J225K080AC	2.2 μF	6.3 V	1.6 mm × 0.8 mm × 0.8 mm	TDK Corporation
C _{REG}	CGA3E1X7R1C105K080AC	1 μF	16 V	1.6 mm × 0.8 mm × 0.8 mm	TDK Corporation
L ₁ , L ₂	TFM201610ALMA2R2MTAA	2.2 μH	20 V	2.0 mm × 1.6 mm × 1.0 mm	TDK Corporation

4.5.1 V_{OUT1} = 5.0 V

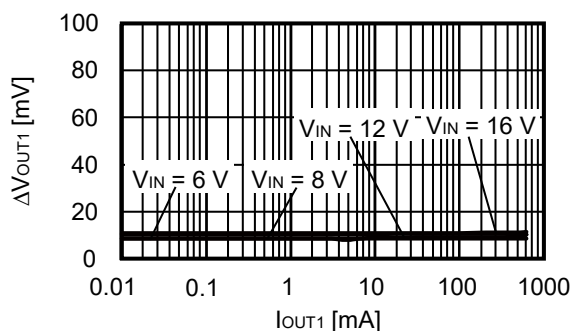
4.5.1.1 Efficiency (η_1) vs Output current (I_{OUT1})

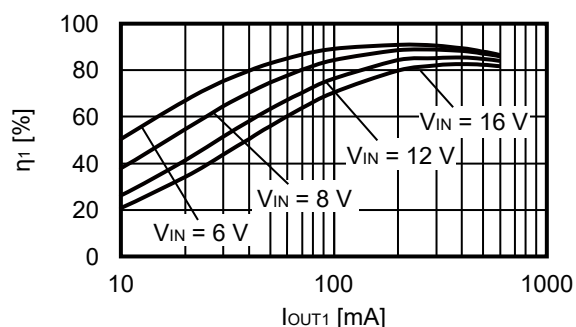
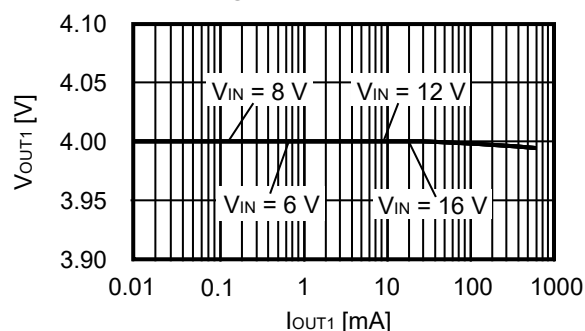
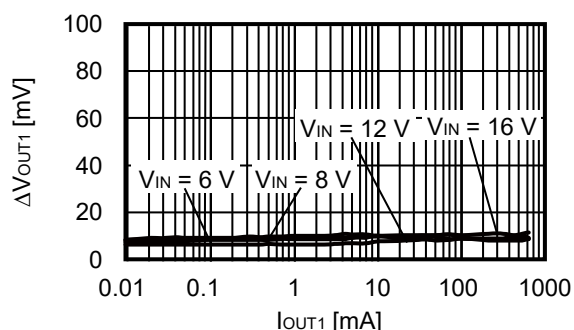
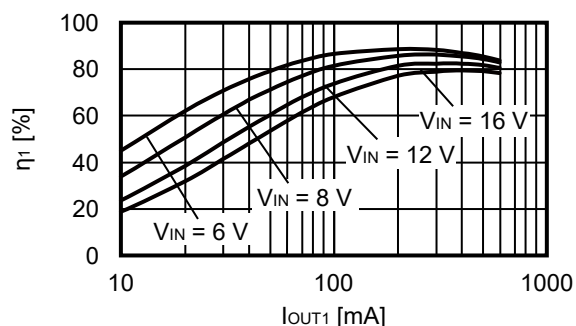
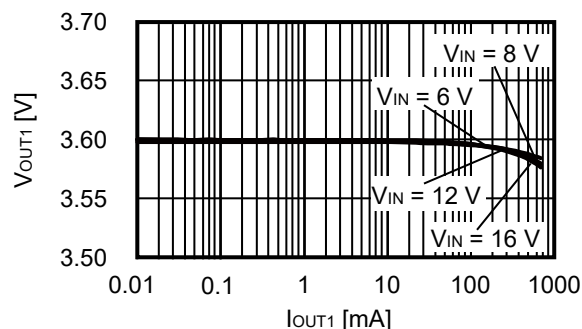
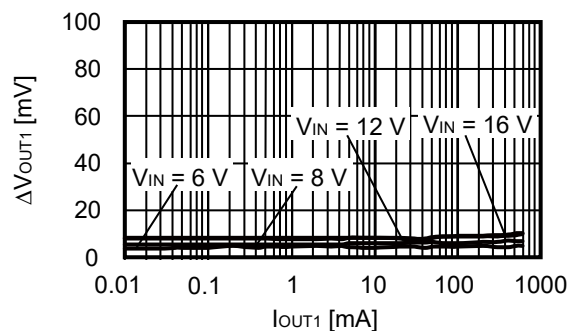


4.5.1.2 Output voltage (V_{OUT1}) vs Output current (I_{OUT1})



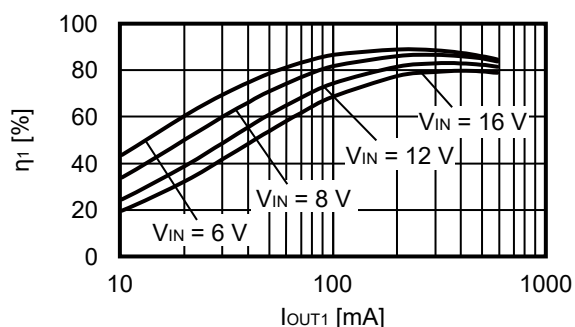
4.5.1.3 Ripple voltage (ΔV_{OUT1}) vs Output current (I_{OUT1})



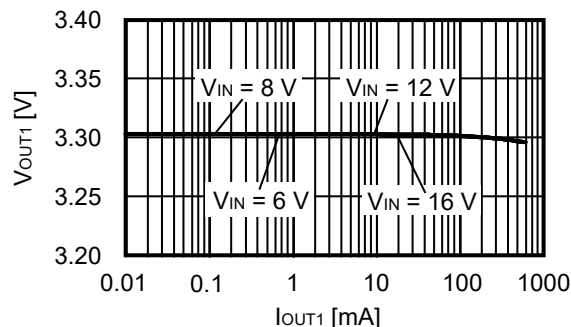
4. 5. 2 $V_{OUT1} = 4.0\text{ V}$ **4. 5. 2. 1 Efficiency (η_1) vs Output current (I_{OUT1})****4. 5. 2. 2 Output voltage (V_{OUT1}) vs Output current (I_{OUT1})****4. 5. 2. 3 Ripple voltage (ΔV_{OUT1}) vs Output current (I_{OUT1})****4. 5. 3 $V_{OUT1} = 3.6\text{ V}$** **4. 5. 3. 1 Efficiency (η_1) vs Output current (I_{OUT1})****4. 5. 3. 2 Output voltage (V_{OUT1}) vs Output current (I_{OUT1})****4. 5. 3. 3 Ripple voltage (ΔV_{OUT1}) vs Output current (I_{OUT1})**

4.5.4 $V_{OUT1} = 3.3\text{ V}$

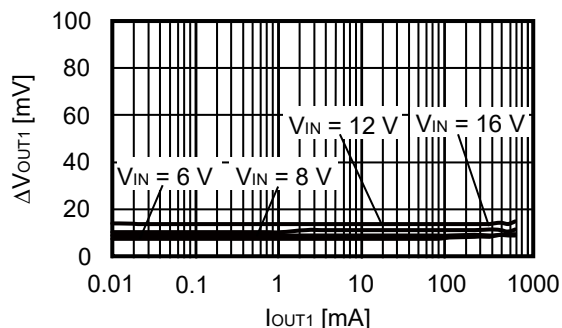
4.5.4.1 Efficiency (η_1) vs Output current (I_{OUT1})



4.5.4.2 Output voltage (V_{OUT1}) vs Output current (I_{OUT1})

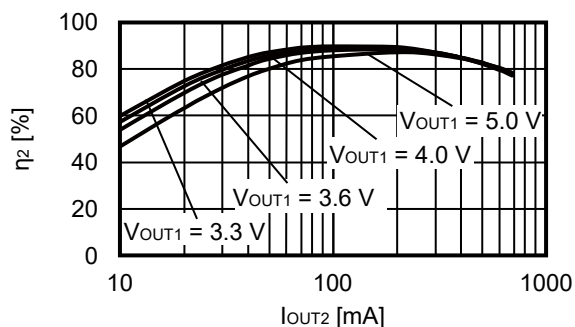


4.5.4.3 Ripple voltage (ΔV_{OUT1}) vs Output current (I_{OUT1})

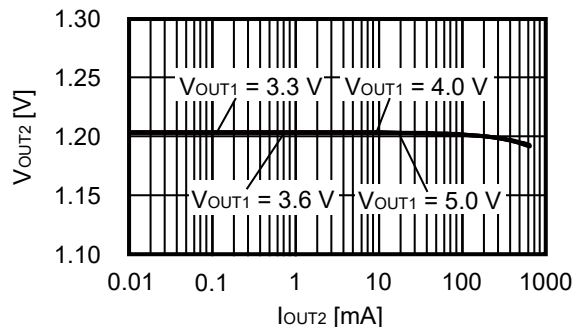


4.5.5 $V_{OUT2} = 1.2\text{ V}$

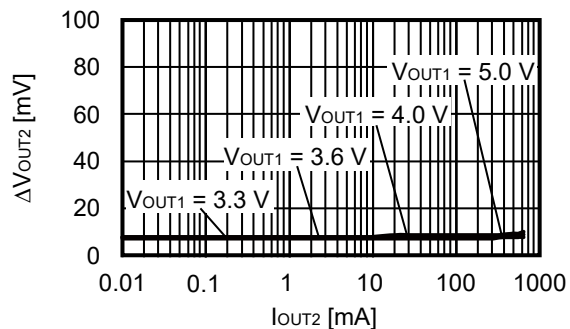
4.5.5.1 Efficiency (η_2) vs Output current (I_{OUT2})

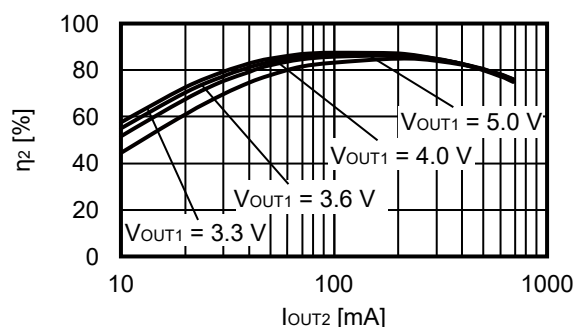
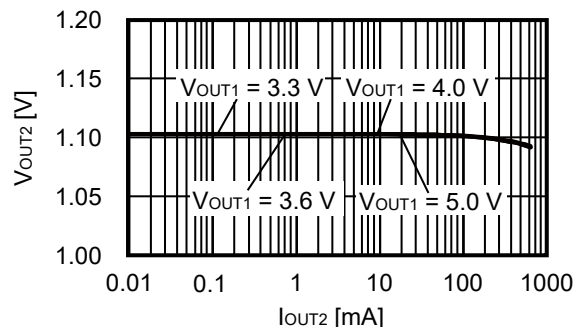
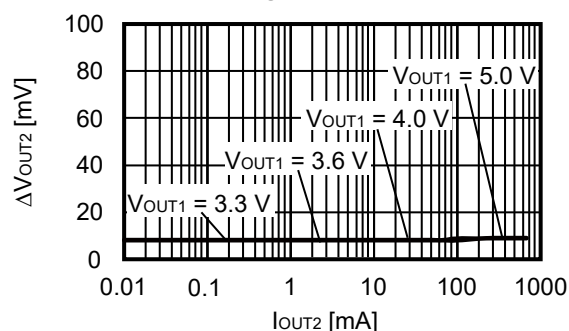
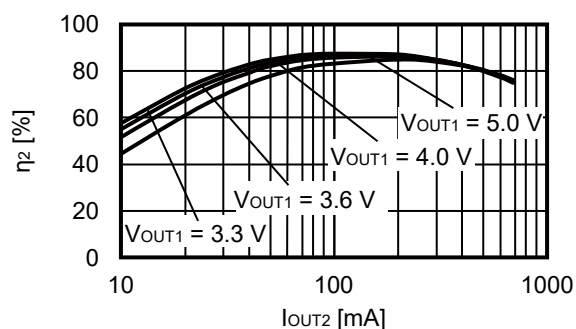
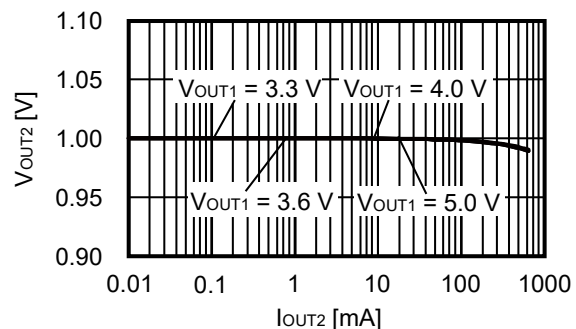
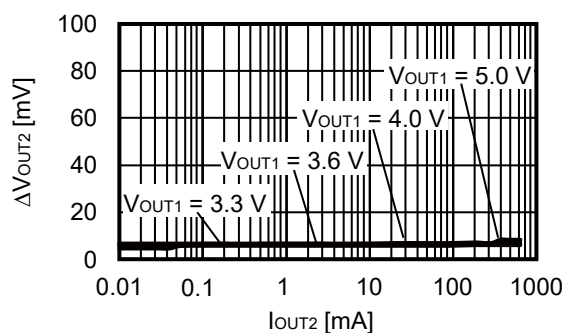


4.5.5.2 Output voltage (V_{OUT2}) vs Output current (I_{OUT2})



4.5.5.3 Ripple voltage (ΔV_{OUT2}) vs Output current (I_{OUT2})



4. 5. 6 $V_{OUT2} = 1.1\text{ V}$ **4. 5. 6. 1 Efficiency (η_2) vs Output current (I_{OUT2})****4. 5. 6. 2 Output voltage (V_{OUT2}) vs Output current (I_{OUT2})****4. 5. 6. 3 Ripple voltage (ΔV_{OUT2}) vs Output current (I_{OUT2})****4. 5. 7 $V_{OUT2} = 1.0\text{ V}$** **4. 5. 7. 1 Efficiency (η_2) vs Output current (I_{OUT2})****4. 5. 7. 2 Output voltage (V_{OUT2}) vs Output current (I_{OUT2})****4. 5. 7. 3 Ripple voltage (ΔV_{OUT2}) vs Output current (I_{OUT2})**

5. Board Layout Precautions

The PCB board layout and component placement are extremely important to ensure stable operation of the DC-DC converter, and achieve regulator performance, conversion efficiency, and noise reduction. This section will explain key points for board layout using a PCB board created by ABLIC Inc.

5.1 Application circuit diagram

Figure 7 shows a PCB board application circuit diagram for the S-19560B Series and S-19255 Series created by ABLIC Inc. The S-19255 Series is connected to the VOUT1 output to generate a low noise analog power supply.

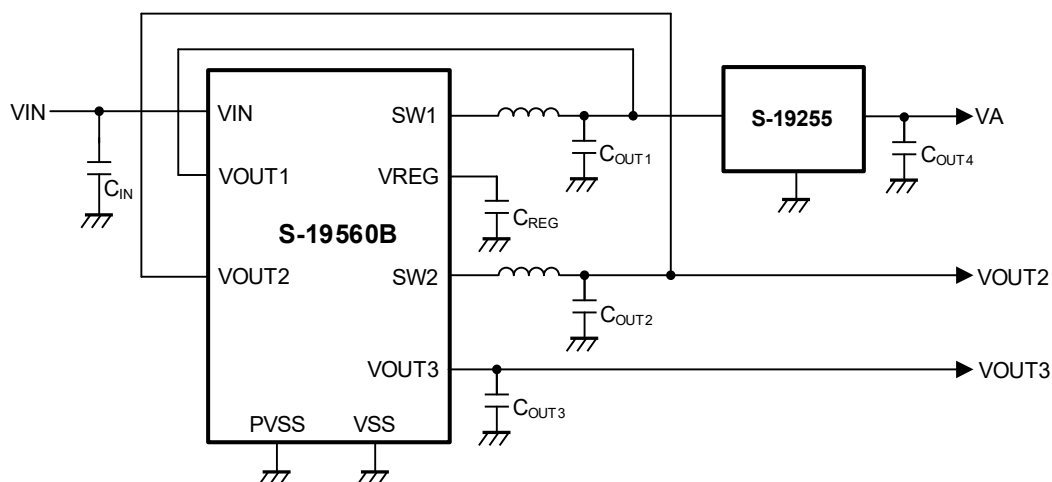


Figure 7

5.2 PCB board layout

The board created by our company for the application circuit in **Figure 7** is a double-sided board. If further heat dissipation is required, use a 4-layer board.

- TOP layer: Layout of most components and wiring
- BOTTOM layer: Most of the area is VSS to improve heat dissipation and reduce noise

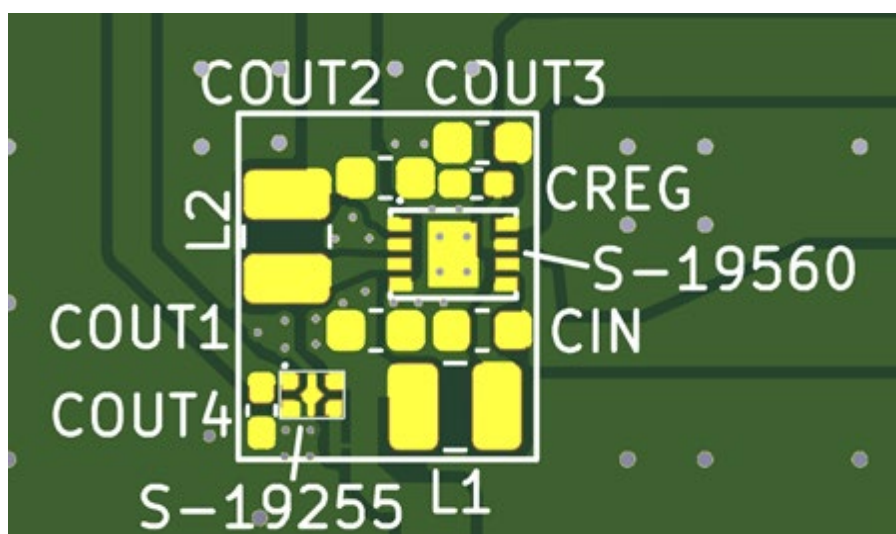


Figure 8 Board Layout Example (Enlarged)

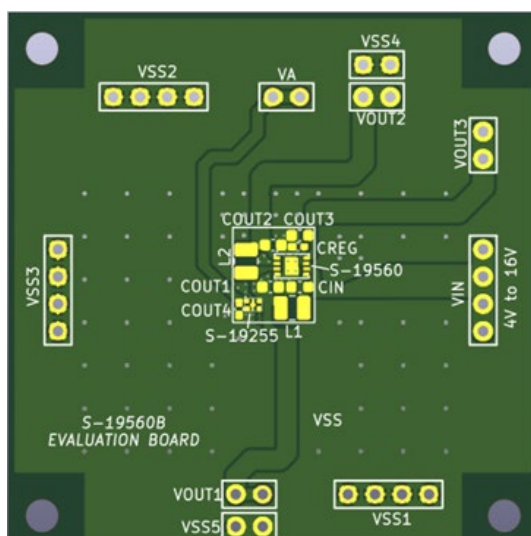


Figure 9 Overall TOP Layer View (Layer 1)

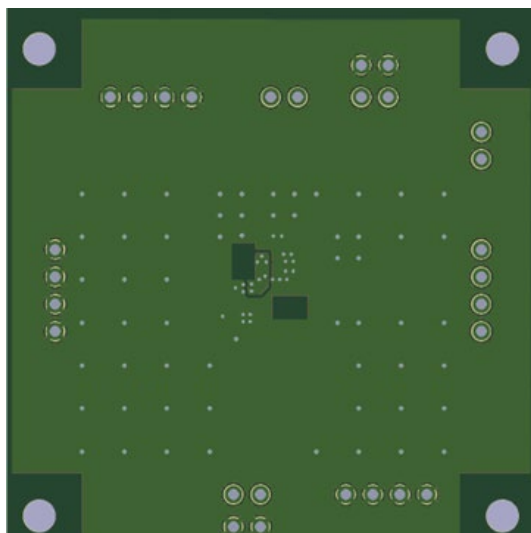


Figure 10 Overall BOTTOM Layer View (Layer 2)

5.3 Board layout precautions

- Place C_{IN} as close to the VIN pin and the VSS pin as possible. Prioritize the layout of C_{IN} .
- Place C_{REG} as close to the VREG pin and the VSS pin as possible.
- Mount C_{IN} and C_{REG} on the same surface layer as the IC. If they are connected through thermal vias, the impedance of the thermal vias may influence the operation, resulting in unstable condition.
- Make the GND pattern as wide as possible.
- Place thermal vias in the GND pattern to ensure sufficient heat dissipation.
- Large current will flow to the SW1 pin and SW2 pin. Make the wiring area of the pattern to be connected to the SW1 pin and SW2 pin small to minimize parasitic capacitance and emission noise.
- Keep the SW1 pin \rightarrow L1 \rightarrow COUT1 \rightarrow VSS pin, SW2 pin \rightarrow L2 \rightarrow COUT2 \rightarrow VSS pin loop wiring short. This is an effective way to reduce emission noise.
- Do not wire the SW1 pin or SW2 pin pattern under the IC.
- In the board layout example in **Figure 8**, the distance between the S-19255 Series input and COUT1 is shortened in order to omit the input capacitor for the S-19255 Series. If the S-19255 Series will be installed in a more distant location, add an input capacitor for the S-19255 Series.

6. Precautions

- The usages described in this application note are typical examples using ICs of ABLIC Inc. Perform thorough evaluation before use.
- Do not apply an electrostatic discharge to this IC that exceeds the performance ratings of the built-in electrostatic protection circuit.
- ABLIC Inc. claims no responsibility for any disputes arising out of or in connection with any infringement by products including this IC of patents owned by a third party.

7. Related Sources

Refer to the following datasheet for details of our products.

S-19560B Series Datasheet

S-19255 Series Datasheet

The information described in this application note and the datasheet is subject to change without notice.

Contact our sales representatives for details.

Regarding the newest version of the datasheet, select product category and product name on our website, and download the PDF file.

www.ablic.com **ABLIC Inc. website**

Disclaimers (Handling Precautions)

1. All the information described herein (product data, specifications, figures, tables, programs, algorithms and application circuit examples, etc.) is current as of publishing date of this document and is subject to change without notice.
2. The circuit examples and the usages described herein are for reference only, and do not guarantee the success of any specific mass-production design.
ABLIC Inc. is not liable for any losses, damages, claims or demands caused by the reasons other than the products described herein (hereinafter "the products") or infringement of third-party intellectual property right and any other right due to the use of the information described herein.
3. ABLIC Inc. is not liable for any losses, damages, claims or demands caused by the incorrect information described herein.
4. Be careful to use the products within their ranges described herein. Pay special attention for use to the absolute maximum ratings, operation voltage range and electrical characteristics, etc.
ABLIC Inc. is not liable for any losses, damages, claims or demands caused by failures and / or accidents, etc. due to the use of the products outside their specified ranges.
5. Before using the products, confirm their applications, and the laws and regulations of the region or country where they are used and verify suitability, safety and other factors for the intended use.
6. When exporting the products, comply with the Foreign Exchange and Foreign Trade Act and all other export-related laws, and follow the required procedures.
7. The products are strictly prohibited from using, providing or exporting for the purposes of the development of weapons of mass destruction or military use. ABLIC Inc. is not liable for any losses, damages, claims or demands caused by any provision or export to the person or entity who intends to develop, manufacture, use or store nuclear, biological or chemical weapons or missiles, or use any other military purposes.
8. The products are not designed to be used as part of any device or equipment that may affect the human body, human life, or assets (such as medical equipment, disaster prevention systems, security systems, combustion control systems, infrastructure control systems, vehicle equipment, traffic systems, in-vehicle equipment, aviation equipment, aerospace equipment, and nuclear-related equipment), excluding when specified for in-vehicle use or other uses by ABLIC, Inc. Do not apply the products to the above listed devices and equipments.
ABLIC Inc. is not liable for any losses, damages, claims or demands caused by unauthorized or unspecified use of the products.
9. In general, semiconductor products may fail or malfunction with some probability. The user of the products should therefore take responsibility to give thorough consideration to safety design including redundancy, fire spread prevention measures, and malfunction prevention to prevent accidents causing injury or death, fires and social damage, etc. that may ensue from the products' failure or malfunction.
The entire system in which the products are used must be sufficiently evaluated and judged whether the products are allowed to apply for the system on customer's own responsibility.
10. The products are not designed to be radiation-proof. The necessary radiation measures should be taken in the product design by the customer depending on the intended use.
11. The products do not affect human health under normal use. However, they contain chemical substances and heavy metals and should therefore not be put in the mouth. The fracture surfaces of wafers and chips may be sharp. Be careful when handling these with the bare hands to prevent injuries, etc.
12. When disposing of the products, comply with the laws and ordinances of the country or region where they are used.
13. The information described herein contains copyright information and know-how of ABLIC Inc. The information described herein does not convey any license under any intellectual property rights or any other rights belonging to ABLIC Inc. or a third party. Reproduction or copying of the information from this document or any part of this document described herein for the purpose of disclosing it to a third-party is strictly prohibited without the express permission of ABLIC Inc.
14. For more details on the information described herein or any other questions, please contact ABLIC Inc.'s sales representative.
15. This Disclaimers have been delivered in a text using the Japanese language, which text, despite any translations into the English language and the Chinese language, shall be controlling.

2.4-2019.07