

## **S-19310 Series**

## AUTOMOTIVE, 125°C OPERATION, 36 V INPUT, 40 mA VOLTAGE REGULATOR WITH RESET FUNCTION

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The S-19310 Series, developed by using high-withstand voltage CMOS process technology, is a positive voltage regulator with the reset function, which has high-withstand voltage and low current consumption.

Regarding a release signal output in the reset function, the S-19310 Series enables delay time adjustment by an external capacitor. Output form of the reset function is selectable from Nch open-drain output or CMOS output.

ABLIC Inc. offers a "thermal simulation service" which supports the thermal design in conditions when our power management ICs are in use by customers. Our thermal simulation service will contribute to reducing the risk in the thermal design at customers' development stage.

ABLIC Inc. also offers FIT rate calculated based on actual customer usage conditions in order to support customer functional safety design.

Contact our sales representatives for details.

Caution This product can be used in vehicle equipment and in-vehicle equipment. Before using the product for these purposes, it is imperative to contact our sales representatives.

#### ■ Features

#### Regulator block

• Output voltage: 2.9 V to 5.3 V, selectable in 0.1 V step

• Input voltage: 3.0 V to 36.0 V

• Output voltage accuracy:  $\pm 2.0\%$  (T<sub>i</sub> = -40°C to +125°C)

Dropout voltage: 240 mV typ. (V<sub>OUT(S)</sub> = 5.0 V, I<sub>OUT</sub> = 30 mA)
 Output current: Possible to output 40 mA (V<sub>IN</sub> = V<sub>OUT(S)</sub> + 2.0 V)\*1
 Input and output capacitors: A ceramic capacitor can be used. (1.0 μF or more)

Built-in overcurrent protection circuit:
 Built-in thermal shutdown circuit:
 Limits overcurrent of output transistor
 Detection temperature 160°C typ.

#### **Detector block**

Detection voltage:
 2.6 V to V<sub>OUT(S)</sub> – 0.3 V, selectable in 0.1 V step

Operation voltage:

 A type: 1.8 V to 36.0 V
 B type: 2.5 V to 36.0 V

 Detection voltage accuracy: ±2.0% (T<sub>j</sub> = -40°C to +125°C)

• Hysteresis width selectable from "Available":  $5.0\% \le V_{HYS} \le 30.0\%$  (T<sub>j</sub> = -40°C to +125°C)

"Available" / "Unavailable": "Unavailable":  $V_{HYS} = 0\%$ 

• Release delay time accuracy:  $\pm 20\%$  (C<sub>D</sub> = 3.3 nF, T<sub>j</sub> = -40°C to +125°C)

Release delay time accuracy: ±20% (C<sub>D</sub> = 3.3 nF, T<sub>j</sub> = -40°C to +125°C<sub>j</sub>
 Output form: Nch open-drain output

Output form:
 Nch open-drain outp
 CMOS output

\_\_

## Overall

• Current consumption: 2.2  $\mu$ A typ. (T<sub>j</sub> = -40°C to +125°C)

• Operation temperature range:  $Ta = -40^{\circ}C$  to  $+125^{\circ}C$ 

• Lead-free (Sn 100%), halogen-free

Withstand 45 V load dump

AEC-Q100 qualified\*2

- \*1. Please make sure that the loss of the IC will not exceed the power dissipation when the output current is large.
- \*2. Contact our sales representatives for details.

## Applications

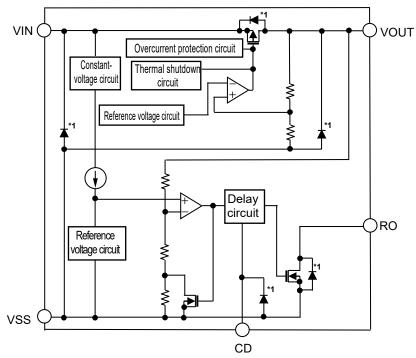
- Constant-voltage power supply and reset circuit for automotive electric component
- Power supply and reset circuit for low-current battery-powered device

#### Packages

- SOT-89-5
- HTMSOP-8
- SOT-23-5

## **■** Block Diagrams

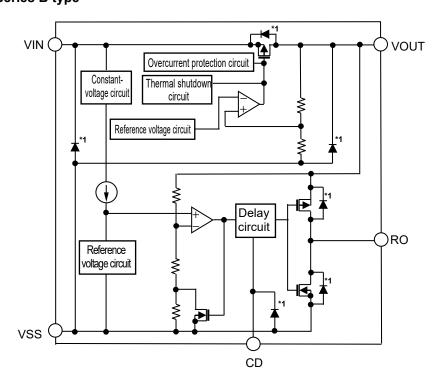
## 1. S-19310 Series A type



\*1. Parasitic diode

Figure 1

## 2. S-19310 Series B type



\*1. Parasitic diode

Figure 2

<sup>2</sup> ABLIC Inc.

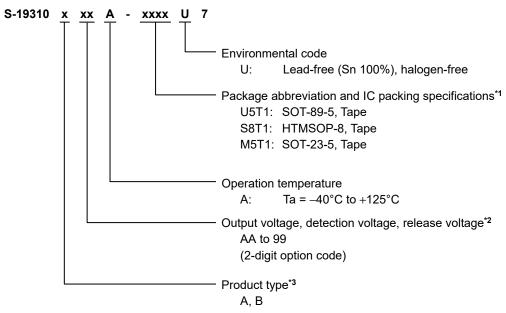
#### ■ AEC-Q100 Qualified

This IC supports AEC-Q100 for operation temperature grade 1. Contact our sales representatives for details of AEC-Q100 reliability specification.

#### **■ Product Name Structure**

Users can select the product type, output voltage, detection voltage, release voltage, and package type for the S-19310 Series. Refer to "1. Product name" regarding the contents of product name, "3. Packages" regarding the package drawings.

#### 1. Product name



- \*1. Refer to the tape drawing.
- **\*2.** Contact our sales representatives for details on combination of output voltage, detection voltage, and release voltage.
- \*3. Refer to "2. Function list of product types".

Remark 1. The output voltage (Vout) can be set in a range which satisfies the following conditions.

- Set output voltage (Vout(S)) is 100 mV step
- $\bullet \ -V_{\text{DET(S)}} + 0.3 \ V \leq V_{\text{OUT(S)}} \leq 5.3 \ V$
- 2. The detection voltage (-V<sub>DET</sub>) can be set in a range which satisfies the following conditions.
  - Set detection voltage (-V<sub>DET(S)</sub>) is 100 mV step
  - 2.6  $V \le -V_{DET(S)} \le V_{OUT(S)} 0.3 V$
- **3.** The release voltage (+V<sub>DET</sub>) can be set in a range which satisfies the following conditions. Release voltage possible setting range is shown in **Figure 3**.
  - Set release voltage (+V<sub>DET(S)</sub>) is 100 mV step
  - $5.0\% \le V_{HYS} \le 30.0\%$
  - $V_{OUT(S)} \times 0.98 45 \text{ mV} > +V_{DET(S)} \times 1.02$

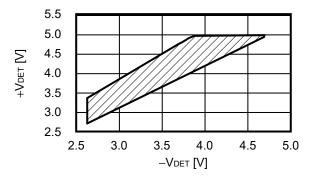


Figure 3 Release Voltage Possible Setting Area

When the difference between  $V_{\text{OUT}}$  and  $+V_{\text{DET}}$  is small, S-19310 Series may not transition from the detection status to the release status due to fluctuation in  $V_{\text{OUT}}$ . Please take transient response characteristics into account when deciding voltages.

If hysteresis width "Unavailable" was selected,  $+V_{DET} = -V_{DET}$ .

## 2. Function list of product types

Table 1

Product Type	RO Pin Output Form	RO Pin Output Logic
Α	Nch open-drain output	Active "L"
В	CMOS output	Active "L"

#### 3. Packages

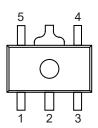
Table 2 Package Drawing Codes

Package Name	Dimension Tape Re		Reel	Land
SOT-89-5	UP005-A-P-SD	UP005-A-C-SD	UP005-A-R-SD	_
HTMSOP-8	FP008-A-P-SD	FP008-A-C-SD	FP008-A-R-SD	FP008-A-L-SD
SOT-23-5	MP005-A-P-SD	MP005-A-C-SD	MP005-A-R-SD	_

## **■** Pin Configurations

#### 1. SOT-89-5

Top view



Pin No.	Symbol	Description
1	CD	Connection pin for delay time adjustment capacitor
2	VSS	GND pin
3	VIN	Input voltage pin
4	VOUT	Output voltage pin
5	RO	Reset output pin

Table 3

Figure 4

Top view

#### 2. HTMSOP-8

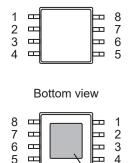


Figure 5

#### Table 4

10010 1						
Pin No.	Symbol	Description				
1	NC*2	No connection				
2	VOUT	Output voltage pin				
3	RO	Reset output pin				
4	NC*2	No connection				
5	CD	Connection pin for delay time adjustment capacitor				
6	VSS	GND pin				
7	NC*2	No connection				
8	VIN	Input voltage pin				

- \*1. Connect the heat sink of backside at shadowed area to the board, and set electric potential GND. However, do not use it as the function of electrode.
- \*2. The NC pin is electrically open. The NC pin can be connected to the VIN pin or the VSS pin.

## 3. SOT-23-5

Top view

Figure 6

Table 5

Pin No.	Symbol	Description
1	RO	Reset output pin
2	VSS	GND pin
3	VOUT	Output voltage pin
4	VIN	Input voltage pin
5 CD		Connection pin for delay time

## ■ Absolute Maximum Ratings

Table 6

(Ta = +25°C unless otherwise specified)

Item		Symbol	Absolute Maximum Rating	Unit
Input voltage		VIN	$V_{SS} - 0.3$ to $V_{SS} + 45.0$	V
Output voltage		Vout	$V_{SS} - 0.3$ to $V_{IN} + 0.3 \le V_{SS} + 7.0$	V
DO nin voltage	Nch open-drain output	\ <u>\</u>	$V_{SS} - 0.3$ to $V_{SS} + 45.0$	V
RO pin voltage	CMOS output	V <sub>RO</sub>	$V_{SS} - 0.3 \text{ to } V_{OUT} + 0.3 \le V_{SS} + 7.0$	V
CD pin voltage		V <sub>CD</sub>	$V_{SS} - 0.3 \text{ to } V_{IN} + 0.3 \leq V_{SS} + 7.0$	V
Output current (Reg	gulator block)	Гоит	52	mA
O. 14m. 14 a. 11ma m4	Nch open-drain output	I <sub>RON</sub>	20	mA
Output current (Detector block)	CNACCtt	Iron	20	mA
(Detector block)	CMOS output	I <sub>ROP</sub>	-20	mA
Junction temperature		Tj	-40 to +150	°C
Operation ambient temperature		Topr	-40 to +125	°C
Storage temperatur	re	T <sub>stg</sub>	-40 to +150	°C

Caution The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

## **■** Thermal Resistance Value

Table 7

Item	Symbol	Conditi	on	Min.	Тур.	Max.	Unit
			Board A	1	119	1	°C/W
			Board B	1	84	1	°C/W
		SOT-89-5	Board C	_	_	_	°C/W
			Board D	_	46	_	°C/W
	Board E - 35	35	_	°C/W			
		JJA HTMSOP-8	Board A	_	159	_	°C/W
	θја		Board B	_	113	_	°C/W
Junction-to-ambient thermal resistance*1			Board C	_	39	_	°C/W
			Board D	_	40	_	°C/W
			Board E	_	30	_	°C/W
			Board A	_	192	_	°C/W
			Board B	_	160	_	°C/W
		SOT23-5	Board C	_	_	_	°C/W
			Board D		_	-	°C/W
			Board E	_	_	_	°C/W

<sup>\*1.</sup> Test environment: compliance with JEDEC STANDARD JESD51-2A

**Remark** Refer to "■ **Power Dissipation**" and "**Test Board**" for details.

#### **■** Electrical Characteristics

#### 1. Regulator block

Table 8

 $(T_i = -40^{\circ}C \text{ to } +125^{\circ}C \text{ unless otherwise specified})$ 

Item	Symbol	Condition		Min.	Тур.	Max.	Unit	Test Circuit
Output voltage*1	V <sub>OUT(E)</sub>	$V_{IN} = V_{OUT(S)} + 2$	2.0 V, I <sub>OUT</sub> = 1 mA	$V_{OUT(S)} \times 0.98$	V <sub>OUT(S)</sub>	$V_{OUT(S)} \times 1.02$	٧	1
Output current*2	Іоит	$V_{IN} = V_{OUT(S)} + 2$	.0 V	40*4	_	_	mA	2
			$2.9 \text{ V} \le V_{OUT(S)} < 3.0 \text{ V}$	0.10	0.39	0.65	>	1
Dropout voltage*3	V <sub>drop</sub>	I <sub>OUT</sub> = 30 mA	$3.0 \text{ V} \le V_{OUT(S)} < 4.0 \text{ V}$	_	0.35	0.60	>	1
			$4.0 \text{ V} \leq V_{OUT(S)} \leq 5.3 \text{ V}$	_	0.24	0.45	>	1
Line regulation	$\frac{\Delta V_{\text{OUT1}}}{\Delta V_{\text{IN}} \bullet V_{\text{OUT}}}$	$V_{OUT(S)} + 2.0 \text{ V} \le V_{IN} \le 36.0 \text{ V}, I_{OUT} = 1 \text{ mA}$		_	0.01	0.2	%/V	1
Load regulation	$\Delta V_OUT2$	$V_{IN} = V_{OUT(S)} + 2$	$1.0 \text{ V}, 1 \mu\text{A} \le \text{I}_{\text{OUT}} \le 30 \text{ mA}$	_	24	45	mV	1
Input voltage	Vin		_	3.0	_	36.0	V	_
Short-circuit current	Ishort	$V_{IN} = V_{OUT(S)} + 2.0 \text{ V}, V_{OUT} = 0 \text{ V}$		_	24	-	mA	2
Thermal shutdown detection temperature	T <sub>SD</sub>	Junction temperature		_	160	_	°C	_
Thermal shutdown release temperature	T <sub>SR</sub>	Junction temper	ature	_	135	_	°C	_

<sup>\*1.</sup> V<sub>OUT(S)</sub>: Set output voltage

 $V_{\text{IN1}}$  is the input voltage at which the output voltage becomes 98% of  $V_{\text{OUT3}}$  after gradually decreasing the input voltage.

 $V_{\text{OUT3}}$  is the output voltage when  $V_{\text{IN}}$  =  $V_{\text{OUT(S)}}$  + 2.0 V, and  $I_{\text{OUT}}$  = 30 mA

\*4. Due to limitation of the power dissipation, this value may not be satisfied. Attention should be paid to the power dissipation when the output current is large.

This specification is guaranteed by design.

V<sub>OUT(E)</sub>: Actual output voltage

Output voltage when fixing  $I_{OUT}$  (= 1 mA) and inputting  $V_{OUT(S)}$  + 2.0 V

<sup>\*2.</sup> The output current at which the output voltage becomes 95% of V<sub>OUT(E)</sub> after gradually increasing the output current.

<sup>\*3.</sup>  $V_{drop} = V_{IN1} - (V_{OUT3} \times 0.98)$ 

#### 2. Detector block

#### Table 9

 $(T_i = -40^{\circ}C \text{ to } +125^{\circ}C \text{ unless otherwise specified})$ 

Item	Symbol	Condition	Condition		Тур.	Max.	Unit	Test Circuit
Detection voltage*1	-V <sub>DET</sub>	_		-V <sub>DET(S)</sub> × 0.98	-V <sub>DET(S)</sub>	-V <sub>DET(S)</sub> × 1.02	٧	3
Delegas valtage*2	11/	V <sub>HYS</sub> = 0%		$-V_{DET(S)} \times 0.98$	-V <sub>DET(S)</sub>	$-V_{DET(S)} \times 1.02$	٧	3
Release voltage*2 +VDE		5.0% ≤ V <sub>HYS</sub> ≤ 30.0%		+V <sub>DET(S)</sub> × 0.98	+V <sub>DET(S)</sub>	+V <sub>DET(S)</sub> × 1.02	٧	3
On anation walters		S-19310 Series A type		1.8	_	36.0	V	_
Operation voltage	Vopr	S-19310 Series B type		2.5	_	36.0	V	
Output current "H"	I <sub>ROP</sub>	CMOS output, $V_{DS}^{*3} = 0.5 \text{ V}$	V <sub>IN</sub> = 16.0 V	-	_	-0.68	mA	4
		CMOS output, $V_{DS}^{*3} = 0.5 \text{ V}$	$V_{IN} = 2.5 V$	3.00	_	_	mA	4
Output current "L"	I <sub>RON</sub>	Nch open-drain output, $V_{DS}^{*3} = 0.5 \text{ V}$	V <sub>IN</sub> = 2.5 V	2.80	_	_	mA	4
Leakage current	ILEAKN	Nch open-drain output	V <sub>RO</sub> = 36.0 V, V <sub>IN</sub> = 36.0 V	_	_	2.0	μА	4
Detection response time*4	t <sub>RESET</sub>	_		_	140	_	μs	5
Release delay time*5	t <sub>DELAY</sub>	C <sub>D</sub> = 3.3 nF		8.0	10.0	12.0	ms	5

- \*1.  $-V_{DET(S)}$ : Set detection voltage,  $-V_{DET}$ : Actual detection voltage
- \*2. +V<sub>DET(S)</sub>: Set release voltage, +V<sub>DET</sub>: Actual release voltage
- \*3. V<sub>DS</sub>: Drain-to-source voltage of the output transistor
- \*4. The time period from when the pulse voltage of  $V_{OUT(S)} \rightarrow 2.5 \text{ V}$  is applied to the VOUT pin within  $2.6 \text{ V} \le -V_{DET(S)} < 3.5 \text{ V}$  or the pulse voltage of  $V_{OUT(S)} \rightarrow -V_{DET(S)} 1.0 \text{ V}$  is applied to the VOUT pin within  $3.5 \text{ V} \le -V_{DET(S)} \le 5.0 \text{ V}$  until  $V_{RO}$  reaches 50% of either  $V_{IN}$  or  $V_{OUT}$
- \*5. The time period from when the pulse voltage of 2.5 V  $\rightarrow$  V<sub>OUT(S)</sub> is applied to the VOUT pin within 2.6 V  $\leq$  +V<sub>DET(S)</sub> < 3.5 V or the pulse voltage of +V<sub>DET(S)</sub> 1.0 V  $\rightarrow$  V<sub>OUT(S)</sub> is applied to the VOUT pin within 3.5 V  $\leq$  +V<sub>DET(S)</sub>  $\leq$  5.0 V until V<sub>RO</sub> reaches 50% of either V<sub>IN</sub> or V<sub>OUT</sub>

#### 3. Overall

Table 10

 $(T_i = -40^{\circ}C \text{ to } +125^{\circ}C \text{ unless otherwise specified})$ 

Item	Symbol	Condition		Min.	Тур.	Max.	Unit	Test Circuit
Command assessmentian	laa.	$V_{IN} = V_{OUT(S)}^{*1} + 2.0 V,$	S-19310 Series A type	-	2.2	4.7	μΑ	6
Current consumption	ISS1	$I_{OUT} = 0 \text{ mA}$	S-19310 Series B type	-	2.2	5.0	μΑ	6

<sup>\*1.</sup> V<sub>OUT(S)</sub>: Set output voltage

#### **■** Test Circuits

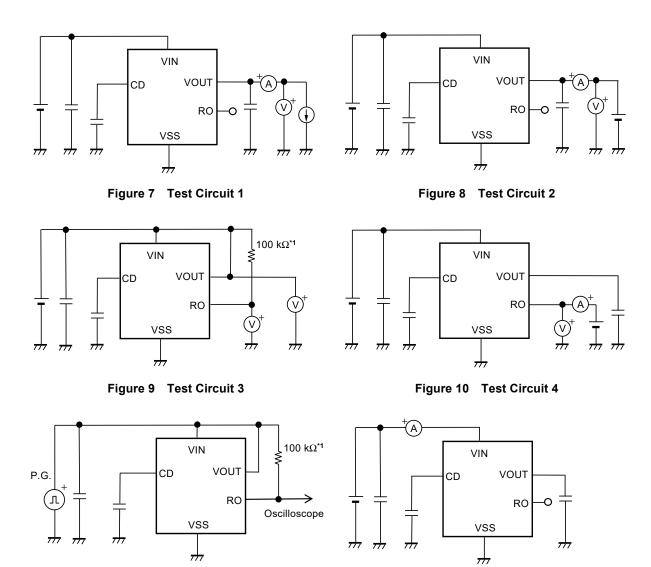


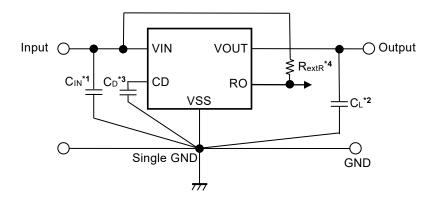
Figure 12 Test Circuit 6

\*1. Only S-19310 Series A type

Figure 11 Test Circuit 5

#### ■ Standard Circuits

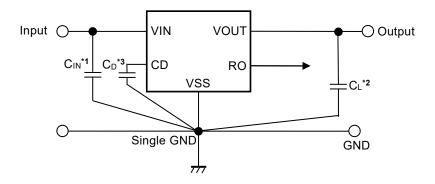
#### 1. S-19310 Series A type



- \*1. C<sub>IN</sub> is a capacitor for stabilizing the input.
- \*2.  $C_L$  is a capacitor for stabilizing the output.
- $\star 3$ .  $C_D$  is the delay time adjustment capacitor.
- \*4. RextR is the external pull-up resistor for the reset output pin.

Figure 13

#### 2. S-19310 Series B type



- \*1. C<sub>IN</sub> is a capacitor for stabilizing the input.
- \*2. C<sub>L</sub> is a capacitor for stabilizing the output.
- ${}^{\star}$ 3. C<sub>D</sub> is the delay time adjustment capacitor.

Figure 14

Caution The above connection diagrams and constants will not guarantee successful operation. Perform thorough evaluation including the temperature characteristics with an actual application to set the constants.

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## ■ Condition of Application

Input capacitor ( $C_{IN}$ ): A ceramic capacitor with capacitance of 1.0  $\mu F$  or more is recommended. Output capacitor ( $C_{L}$ ): A ceramic capacitor with capacitance of 1.0  $\mu F$  or more is recommended. Delay time adjustment capacitor ( $C_{D}$ ): A ceramic capacitor with capacitance of 0.1  $\mu F$  or more is recommended.

Caution Generally, in a voltage regulator, an oscillation may occur depending on the selection of the external parts. Perform thorough evaluation including the temperature characteristics with an actual application using the above capacitors to confirm no oscillation occurs.

## ■ Selection of Input Capacitor (C<sub>IN</sub>) and Output Capacitor (C<sub>L</sub>)

The S-19310 Series requires C<sub>L</sub> between the VOUT pin and the VSS pin for phase compensation.

The operation is stabilized by a ceramic capacitor with capacitance of  $1.0 \mu F$  or more. When using an OS capacitor, a tantalum capacitor or an aluminum electrolytic capacitor, the capacitance also must be  $1.0 \mu F$  or more. However, an oscillation may occur depending on the equivalent series resistance (ESR).

Moreover, the S-19310 Series requires C<sub>IN</sub> between the VIN pin and the VSS pin for a stable operation.

Generally, an oscillaiton may occur when a voltage regulator is used under the conditon that the impedance of the power supply is high.

Note that the output voltage ( $V_{\text{OUT}}$ ) transient characteristics varies depending on the capacitance of  $C_{\text{IN}}$  and  $C_{\text{L}}$  and the value of ESR.

Caution Perform thorough evaluation including the temperature characteristics with an actual application to select C<sub>IN</sub> and C<sub>L</sub>.

## ■ Selection of Delay Time Adjustment Capacitor (CD)

In the S-19310 Series, the delay time adjustment capacitor (C<sub>D</sub>) is necessary between the CD pin and the VSS pin to adjust the release delay time (t<sub>DELAY</sub>) of the detector. Refer to "2. 2 Delay circuit" of "2. Detector block" in "■ Operation" for details.

Caution Perform thorough evaluation including the temperature characteristics with an actual application to select Cn.

## ■ Explanation of Terms

#### 1. Regulator block

#### 1. 1 Low dropout voltage regulator

This is a voltage regulator which made dropout voltage small by its built-in low on-resistance output transistor.

#### 1. 2 Output voltage (Vout)

This voltage is output at an accuracy of  $\pm 2.0\%$  when the input voltage, the output current and the temperature are in a certain condition\*1.

\*1. Differs depending on the product.

Caution If the certain condition is not satisfied, the output voltage may exceed the accuracy range of ±2.0%. Refer to "1. Regulator block" in "■ Electrical Characteristics" and "1. Regulator block" in "■ Characteristics (Typical Data)" for details.

1. 3 Line regulation 
$$\left(\frac{\Delta V_{\text{OUT1}}}{\Delta V_{\text{IN}} \bullet V_{\text{OUT}}}\right)$$

Indicates the dependency of the output voltage against the input voltage. That is, the value shows how much the output voltage changes due to a change in the input voltage after fixing output current constant.

#### 1. 4 Load regulation (ΔV<sub>OUT2</sub>)

Indicates the dependency of the output voltage against the output current. That is, the value shows how much the output voltage changes due to a change in the output current after fixing input voltage constant.

#### 1. 5 Dropout voltage (V<sub>drop</sub>)

Indicates the difference between input voltage ( $V_{IN1}$ ) and the output voltage when the output voltage becomes 98% of the output voltage value ( $V_{OUT3}$ ) at  $V_{IN} = V_{OUT(S)} + 2.0$  V after the input voltage ( $V_{IN}$ ) is decreased gradually.

$$V_{drop} = V_{IN1} - (V_{OUT3} \times 0.98)$$

#### 2. Detector block

#### 2. 1 Detection voltage (-VDET)

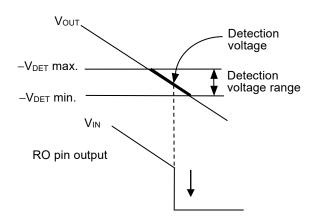
The detection voltage is a voltage at which the output of the RO pin turns to "L".

The detection voltage varies slightly among products of the same specification. The variation of detection voltage between the specified minimum ( $-V_{DET}$  min.) and the maximum ( $-V_{DET}$  max.) is called the detection voltage range (Refer to **Figure 15**, **Figure 17**).

#### 2. 2 Release voltage (+VDET)

The release voltage is a voltage at which the output of the RO pin turns to "H".

The release voltage varies slightly among products of the same specification. The variation of release voltage between the specified minimum (+V<sub>DET</sub> min.) and the maximum (+V<sub>DET</sub> max.) is called the release voltage range (Refer to **Figure 16**, **Figure 18**).



Release voltage
+V<sub>DET</sub> max.
+V<sub>DET</sub> min.

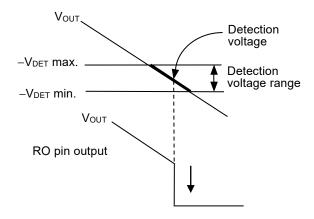
Release voltage range

Vout

Release voltage range

Figure 15 Detection Voltage (S-19310 Series A type)

Figure 16 Release Voltage (S-19310 Series A type)



Release voltage
+VDET max.
+VDET min.

Release voltage range

Vout

Ro pin output

Figure 17 Detection Voltage (S-19310 Series B type)

Figure 18 Release Voltage (S-19310 Series B type)

#### 2. 3 Hysteresis width (V<sub>HYS</sub>)

The hysteresis width is the voltage difference between the detection voltage and the release voltage. Setting the hysteresis width between the detection voltage and the release voltage prevents malfunction caused by noise on the pin voltage in the detection status.

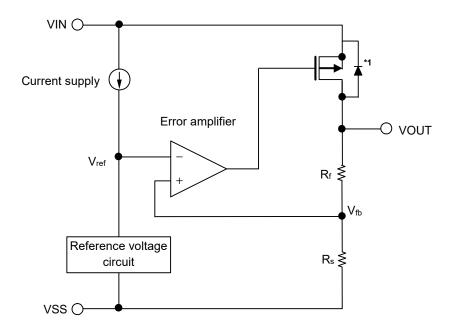
## ■ Operation

#### 1. Regulator block

#### 1. 1 Basic operation

Figure 19 shows the block diagram of the regulator block to describe the basic operation.

The error amplifier compares the feedback voltage ( $V_{fb}$ ) whose output voltage ( $V_{OUT}$ ) is divided by the feedback resistors ( $R_s$  and  $R_f$ ) with the reference voltage ( $V_{ref}$ ). The error amplifier controls the output transistor, consequently, the regulator starts the operation that keeps  $V_{OUT}$  constant without the influence of the input voltage ( $V_{IN}$ ).



\*1. Parasitic diode

Figure 19

#### 1. 2 Output transistor

In the S-19310 Series, a low on-resistance P-channel MOS FET is used between the VIN pin and the VOUT pin as the output transistor. In order to keep  $V_{OUT}$  constant, the on-resistance of the output transistor varies appropriately according to the output current ( $I_{OUT}$ ).

Caution Since a parasitic diode exists between the VIN pin and the VOUT pin due to the structure of the transistor, the IC may be damaged by a reverse current if  $V_{\text{OUT}}$  becomes higher than  $V_{\text{IN}}$ . Therefore, be sure that  $V_{\text{OUT}}$  does not exceed  $V_{\text{IN}} + 0.3 \text{ V}$ .

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#### 1. 3 Overcurrent protection circuit

The S-19310 Series has a built-in overcurrent protection circuit to limit the overcurrent of the output transistor. When the VOUT pin is shorted to the VSS pin, that is, at the time of the output short-circuit, the output current is limited to 24 mA typ. due to the overcurrent protection circuit operation. The S-19310 Series restarts regulating when the output transistor is released from the overcurrent status.

Caution This overcurrent protection circuit does not work as for thermal protection. For example, when the output transistor keeps the overcurrent status long at the time of output short-circuit or due to other reasons, pay attention to the conditions of the input voltage and the load current so as not to exceed the power dissipation.

#### 1. 4 Thermal shutdown circuit

The S-19310 Series has a built-in thermal shutdown circuit to limit overheating. When the junction temperature increases to 160°C typ., the thermal shutdown circuit becomes the detection status, and the regulating is stopped. When the junction temperature decreases to 135°C typ., the thermal shutdown circuit becomes the release status, and the regulator is restarted.

If the thermal shutdown circuit becomes the detection status due to self-heating, the regulating is stopped and  $V_{OUT}$  decreases. For this reason, the self-heating is limited and the temperature of the IC decreases. The thermal shutdown circuit becomes release status when the temperature of the IC decreases, and the regulating is restarted thus the self-heating is generated again. Repeating this procedure makes the waveform of  $V_{OUT}$  into a pulse-like form. This phenomenon continues unless decreasing either or both of the input voltage and the output current in order to reduce the internal power consumption, or decreasing the ambient temperature. Note that the product may suffer physical damage such as deterioration if the above phenomenon occurs continuously.

#### Caution

If a large load current flows during the restart process of regulating after the thermal shutdown circuit changes to the release status from the detection status, the thermal shutdown circuit becomes the detection status again due to self-heating, and a problem may happen in the restart of regulating. A large load current, for example, occurs when charging to the C<sub>L</sub> whose capacitance is large.

Perform thorough evaluation including the temperature characteristics with an actual application to select  $C_{\text{L}}$ .

Table 11

Thermal Shutdown Circuit	VOUT Pin Voltage
Release: 135°C typ.*1	Constant value*2
Detection: 160°C typ.*1	Pulled down to Vss*3

<sup>\*1.</sup> Junction temperature

<sup>\*2.</sup> The constant value is output due to the regulating based on the set output voltage value.

<sup>\*3.</sup> The VOUT pin voltage is pulled down to  $V_{SS}$  due to the feedback resistors ( $R_s$  and  $R_f$ ) and a load.

#### 2. Detector block

#### 2. 1 Basic operation

## 2. 1. 1 S-19310 Series A type

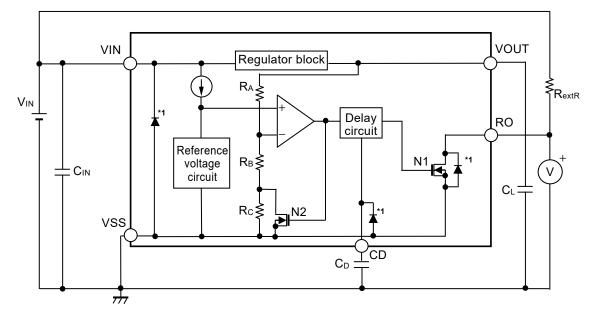
(1) When the output voltage (V<sub>OUT</sub>) of the regulator is release voltage (+V<sub>DET</sub>) of the detector or higher, the Nch transistor (N1) is off, and the RO pin output is "H".

Since the Nch transistor (N2) is off, the input voltage to the comparator is  $\frac{(R_B + R_C) \bullet V_{OUT}}{R_A + R_B + R_C}$ .

(2) Even if V<sub>OUT</sub> decreases to +V<sub>DET</sub> or lower, the RO pin output is "H" when V<sub>OUT</sub> is the detection voltage (-V<sub>DET</sub>) or higher. When V<sub>OUT</sub> is -V<sub>DET</sub> (point A in **Figure 21**) or lower, N1 of output stage is on, and the RO pin output is "L".

At this time, N2 is on, and the input voltage to the comparator is  $\frac{R_B \bullet V_{OUT}}{R_A + R_B}$ .

- (3) When  $V_{IN}$  decreases following  $V_{OUT}$  to the IC's minimum operation voltage or lower, the RO pin output is unstable.
- (4) When  $V_{IN}$  increases and is the IC's minimum operation voltage or higher, the RO pin output is "L". Moreover, even if  $V_{OUT}$  exceeds  $-V_{DET}$ , the RO pin output is "L" when  $V_{OUT}$  is lower than  $+V_{DET}$ .
- (5) When  $V_{OUT}$  further increases to  $+V_{DET}$  (point B in **Figure 21**) or higher, the delay circuit operates. The RO pin output is "H" after the elapse of the release delay time ( $t_{DELAY}$ ).



#### \*1. Parasitic diode

Figure 20 Operation

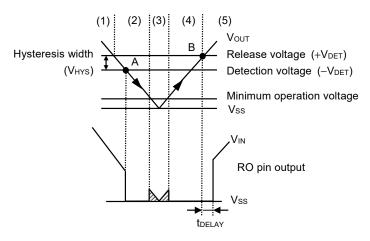
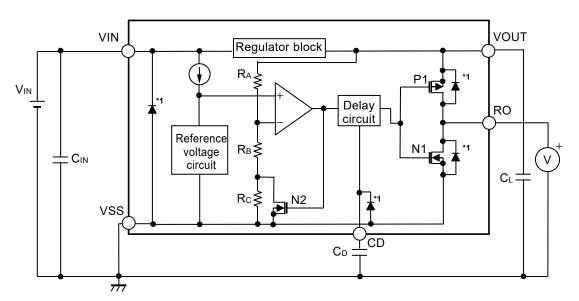


Figure 21 Timing Chart ABLIC Inc.

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#### 2. 1. 2 S-19310 Series B type

- (1) When the output voltage (V<sub>OUT</sub>) of the regulator is release voltage (+V<sub>DET</sub>) of the detector or higher, the Pch transistor (P1) is on, the Nch transistor (N1) is off, and the RO pin output is "H".
  - Since the Nch transistor (N2) is off, the input voltage to the comparator is  $\frac{(R_B + R_C) \bullet V_{OUT}}{R_A + R_B + R_C}$ .
- (2) Even if V<sub>OUT</sub> decreases to +V<sub>DET</sub> or lower, the RO pin output is "H" when V<sub>OUT</sub> is the detection voltage (-V<sub>DET</sub>) or higher. When V<sub>OUT</sub> decreases to -V<sub>DET</sub> (point A in **Figure 23**) or lower, P1 of output stage is off, N1 of output stage is on, and the RO pin output is "L".
  - At this time, N2 is on, and the input voltage to the comparator is  $\frac{R_B \bullet V_{OUT}}{R_A + R_B}$
- (3) When V<sub>IN</sub> decreases following V<sub>OUT</sub> to the IC's minimum operation voltage or lower, the RO pin output is unstable.
- (4) When V<sub>IN</sub> increases to the IC's minimum operation voltage or higher, the RO pin output is "L". Moreover, even if V<sub>OUT</sub> exceeds –V<sub>DET</sub>, the RO pin output is "L" when V<sub>OUT</sub> is lower than +V<sub>DET</sub>.
- (5) When V<sub>OUT</sub> further increases to +V<sub>DET</sub> (point B in **Figure 23**) or higher, the delay circuit operates. The RO pin output is "H" after the elapse of the release delay time (t<sub>DELAY</sub>).



#### \*1. Parasitic diode

Figure 22 Operation

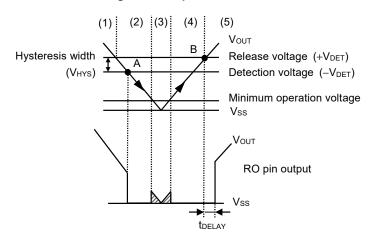


Figure 23 Timing Chart

#### 2. 2 Delay circuit

The delay circuit has a function that adjusts the release delay time ( $t_{DELAY}$ ) from when the output voltage ( $V_{OUT}$ ) of the regulator reaches the release voltage ( $+V_{DET}$ ) of the detector to when the output from the RO pin inverts.  $t_{DELAY}$  is determined by the delay coefficient, the delay time adjustment capacitor ( $C_D$ ) and the release delay time when the CD pin is open ( $t_{DELAY0}$ ). It is calculated by the equation below.

 $t_{DELAY}$  [ms] = Delay coefficient  $\times$  C<sub>D</sub> [nF] +  $t_{DELAY0}$  [ms]

Table 12

lunation Tanananatura	Delay Coefficient				
Junction Temperature	Min.	Тур.	Max.		
$T_j = -40^{\circ}\text{C to } +125^{\circ}\text{C}$	2.76	3.00	3.28		

Table 13

	Release Delay Time			
Junction Temperature	when CD Pin is Open (t <sub>DELAY0</sub> )			
	Min.	Тур.	Max.	
$T_j = -40^{\circ}\text{C to } +125^{\circ}\text{C}$	0.03 ms	0.07 ms	0.36 ms	

- Caution 1. The above equation will not guarantee successful operation. Perform thorough evaluation including the temperature characteristics with an actual application to set the constants.
  - 2. Mounted board layout should be made in such a way that no current flows into or flows from the CD pin since the impedance of the CD pin is high, otherwise correct delay time cannot be provided.
  - 3. There is no limit for the capacitance of C<sub>D</sub> as long as the leakage current of the capacitor can be ignored against the built-in constant current value (approximately 350 nA). The leakage current may cause error in delay time. When the leakage current is larger than the built-in constant current, no detect or release takes place.

#### 2. 3 V<sub>IN</sub> drop during release delay time (reference)

 $t_R = t_F = 10 \, \mu s$ 

Figure 24 shows the relation between pulse width  $(t_{PW})$  and  $V_{IN}$  lower limit  $(V_{DROP})$  where a release signal can be output after the normal release delay time has elapsed when the VIN pin voltage instantaneously drops during release delay time.

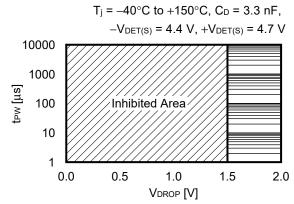


Figure 24

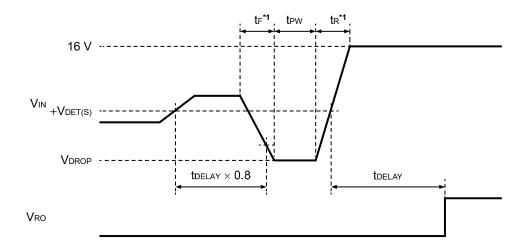


Figure 25 VIN Pin Input Voltage Waveform

- Caution 1. Figure 24 shows the input voltage conditions when a release signal is output after the normal release delay time has elapsed. When this is within the inhibited area, release may erroneously be executed before the delay time completes.
  - 2. When the VIN pin voltage is within the inhibited areas shown in Figure 24 during release delay time, input 0 V to the VIN pin then restart the detector.

#### ■ Precautions

- Generally, when a voltage regulator is used under the condition that the impedance of the power supply is high, an
  oscillation may occur. Perform thorough evaluation including the temperature characteristics with an actual
  application to select C<sub>IN</sub>.
- Generally, in a voltage regulator, an oscillation may occur depending on the selection of the external parts. The following use conditions are recommended in the S-19310 Series; however, perform thorough evaluation including the temperature characteristics with an actual application to select C<sub>IN</sub> and C<sub>L</sub>.

Input capacitor ( $C_{IN}$ ): A ceramic capacitor with capacitance of 1.0  $\mu F$  or more is recommended. Output capacitor ( $C_L$ ): A ceramic capacitor with capacitance of 1.0  $\mu F$  or more is recommended.

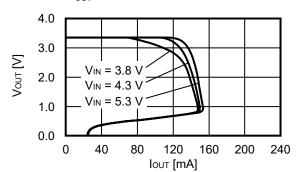
- Generally, in a voltage regulator, the values of an overshoot and an undershoot in the output voltage vary depending on the variation factors of input voltage start-up, input voltage fluctuation, load fluctuation etc., or the capacitance of C<sub>IN</sub> or C<sub>L</sub> and the value of the equivalent series resistance (ESR), which may cause a problem to the stable operation. Perform thorough evaluation including the temperature characteristics with an actual application to select C<sub>IN</sub> and C<sub>L</sub>.
- Generally, in a voltage regulator, an overshoot may occur in the output voltage momentarily if the input voltage steeply changes when the input voltage is started up, the input voltage fluctuates, etc. Perform thorough evaluation including the temperature characteristics with an actual application to confirm no problems happen.
- Generally, in a voltage regulator, if the VOUT pin is steeply shorted with GND, a negative voltage exceeding the
  absolute maximum ratings may occur in the VOUT pin due to resonance phenomenon of the inductance and the
  capacitance including C<sub>L</sub> on the application. The resonance phenomenon is expected to be weakened by inserting
  a series resistor into the resonance path, and the negative voltage is expected to be limited by inserting a
  protection diode between the VOUT pin and the VSS pin.
- If the input voltage is started up steeply under the condition that the capacitance of  $C_L$  is large, the thermal shutdown circuit may be in the detection status by self-heating due to the charge current to  $C_L$ .
- Make sure of the conditions for the input voltage, output voltage and the load current so that the internal loss does not exceed the power dissipation.
- Do not apply an electrostatic discharge to this IC that exceeds the performance ratings of the built-in electrostatic protection circuit.
- When considering the output current value that the IC is able to output, make sure of the output current value specified in Table 8 in "

  Electrical Characteristics" and footnote \*4 of the table.
- Wiring patterns on the application related to the VIN pin, the VOUT pin and the VSS pin should be designed so that the impedance is low. When mounting C<sub>IN</sub> between the VIN pin and the VSS pin and C<sub>L</sub> between the VOUT pin and the VSS pin, connect the capacitors as close as possible to the respective destination pins of the IC.
- In the package equipped with heat sink of backside, mount the heat sink firmly. Since the heat radiation differs according to the condition of the application, perform thorough evaluation with an actual application to confirm no problems happen.
- ABLIC Inc. claims no responsibility for any disputes arising out of or in connection with any infringement by products including this IC of patents owned by a third party.

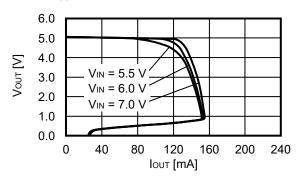
## ■ Characteristics (Typical Data)

## 1. Regulator block

1. 1 Output voltage vs. Output current (When load current increases) (Ta = +25°C)



1. 1. 2 V<sub>OUT</sub> = 5.0 V

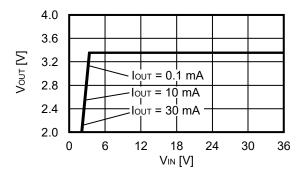


**Remark** In determining the output current, attention should be paid to the following.

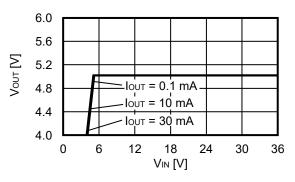
- 1. The minimum output current value and footnote \*4 of Table 8 in "■ Electrical Characteristics"
- 2. Power dissipation

#### 1. 2 Output voltage vs. Input voltage (Ta = +25°C)

1. 2. 1 V<sub>OUT</sub> = 3.3 V

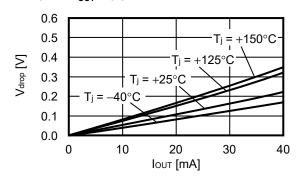


1. 2. 2 V<sub>OUT</sub> = 5.0 V

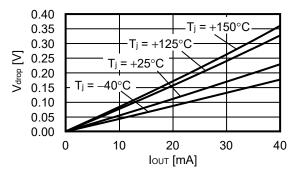


#### 1. 3 Dropout voltage vs. Output current

1. 3. 1 V<sub>OUT</sub> = 3.3 V

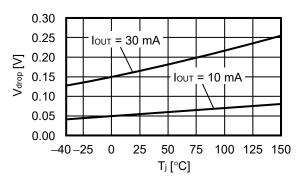


1. 3. 2 V<sub>OUT</sub> = 5.0 V

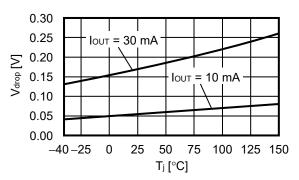


#### 1. 4 Dropout voltage vs. Junction temperature

#### 1. 4. 1 Vout = 3.3 V

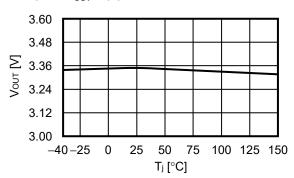


#### 1. 4. 2 V<sub>OUT</sub> = 5.0 V

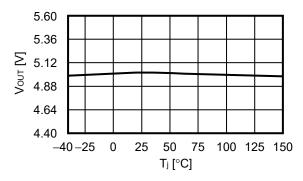


#### 1. 5 Output voltage vs. Junction temperature

1. 5. 1 V<sub>OUT</sub> = 3.3 V

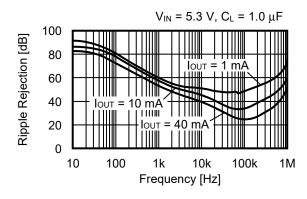


#### 1. 5. 2 V<sub>OUT</sub> = 5.0 V

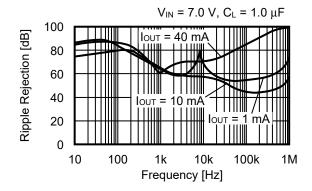


#### 1. 6 Ripple rejection (Ta = +25°C)

#### 1. 6. 1 $V_{OUT} = 3.3 V$

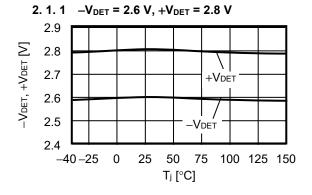


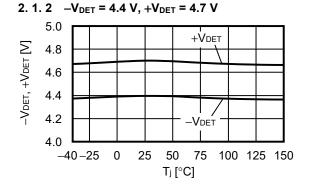
#### 1. 6. 2 V<sub>OUT</sub> = 5.0 V



#### 2. Detector block

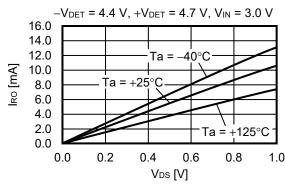
#### 2. 1 Detection voltage, Release voltage vs. Junction temperature



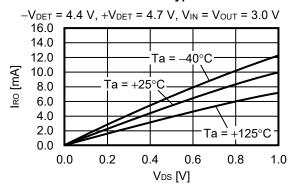


#### 2. 2 Nch transistor output current vs. V<sub>DS</sub>

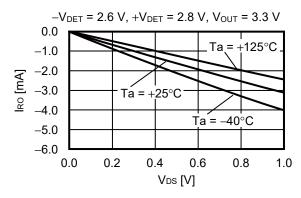
#### 2. 2. 1 S-19310 Series A type



#### 2. 2. 2 S-19310 Series B type



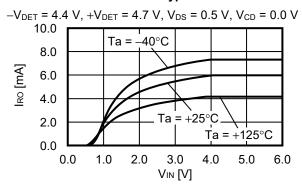
#### 2. 3 Pch transistor output current vs. V<sub>DS</sub> (S-19310 Series B type)



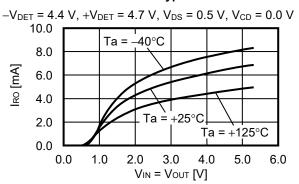
Remark V<sub>DS</sub>: Drain-to-source voltage of the output transistor

#### 2. 4 Nch transistor output current vs. Power supply voltage

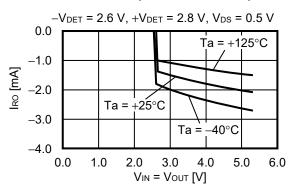
#### 2. 4. 1 S-19310 Series A type



#### 2. 4. 2 S-19310 Series B type



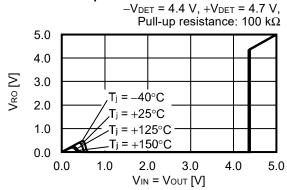
#### 2. 5 Pch transistor output current vs. Output voltage (S-19310 Series B type)



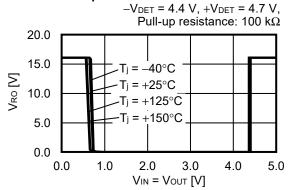
Remark V<sub>DS</sub>: Drain-to-source voltage of the output transistor

## 2. 6 Minimum operation voltage vs. Power supply voltage / Output voltage (S-19310 Series A type)

#### 2. 6. 1 Pull-up to $V_{\text{IN}}$



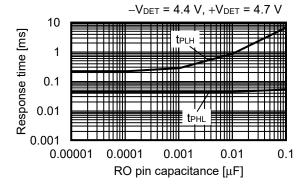
#### 2. 6. 2 Pull-up to 16.0 V



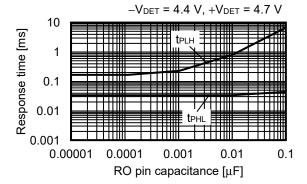
#### 2. 7 Dynamic response vs. RO pin capacitance

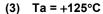
#### 2. 7. 1 S-19310 Series A type

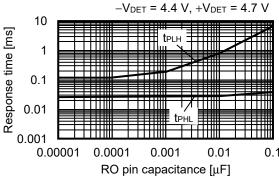
#### (1) $Ta = -40^{\circ}C$

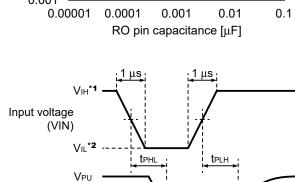




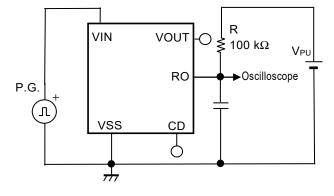








**V**PU × **50%** 



\*1. V<sub>IH</sub> = 36.0 V

(RO)

Output voltage

**\*2.** V<sub>IL</sub> = 2.5 V

Figure 26 Test Condition of Response Time

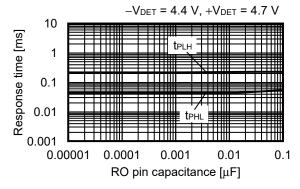
Figure 27 Test Circuit of Response Time

Caution The above connection diagram and constants will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constants.

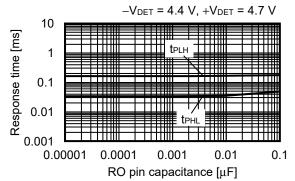
**V**PU × 50%

#### 2. 7. 2 S-19310 Series B type

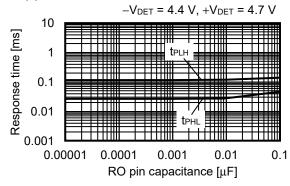
#### (1) $Ta = -40^{\circ}C$

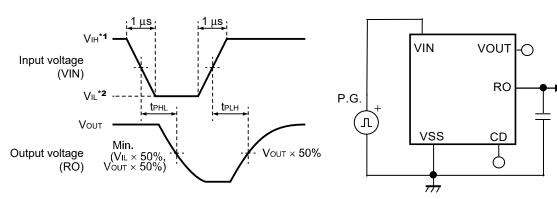






## (3) Ta = +125°C





- \*1. V<sub>IH</sub> = 36.0 V
- \*2. V<sub>IL</sub> = 2.5 V

Figure 28 Test Condition of Response Time

Figure 29 Test Circuit of Response Time

Oscilloscope

Caution The above connection diagram and constants will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constants.

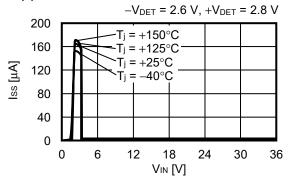
26

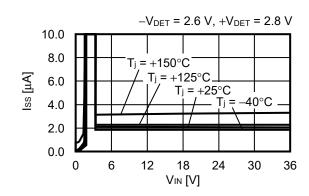
#### 3. Overall

#### 3. 1 Current consumption vs. Input voltage (No load)

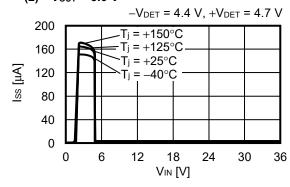
#### 3. 1. 1 S-19310 Series A type

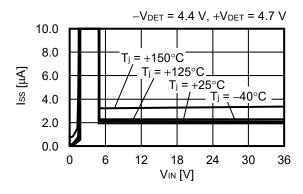
#### (1) $V_{OUT} = 3.3 V$





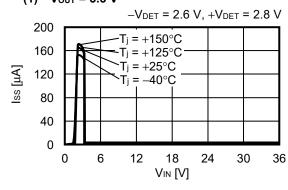
## (2) $V_{OUT} = 5.0 \text{ V}$

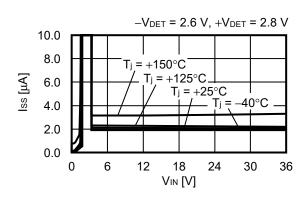




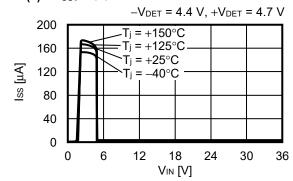
#### 3. 1. 2 S-19310 Series B type

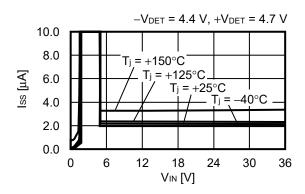
## (1) $V_{OUT} = 3.3 V$





#### (2) $V_{OUT} = 5.0 V$

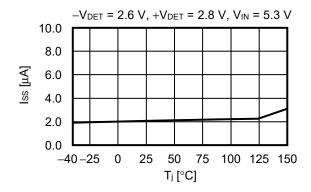




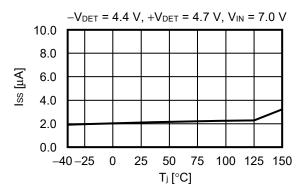
#### 3. 2 Current consumption vs. Junction temperature

#### 3. 2. 1 S-19310 Series A type

#### (1) $V_{OUT} = 3.3 V$

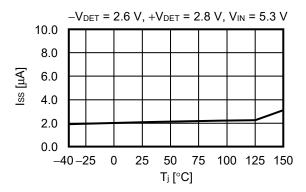


#### (2) $V_{OUT} = 5.0 V$

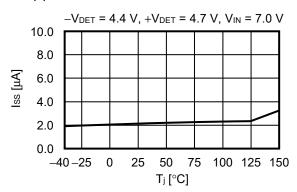


#### 3. 2. 2 S-19310 Series B type

#### (1) $V_{OUT} = 3.3 V$



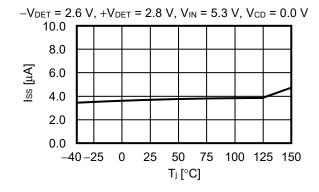
#### (2) $V_{OUT} = 5.0 V$



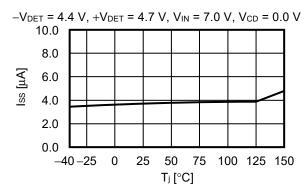
#### 3. 3 Current consumption during release delay time vs. Junction temperature

#### 3. 3. 1 S-19310 Series A type

#### (1) $V_{OUT} = 3.3 V$

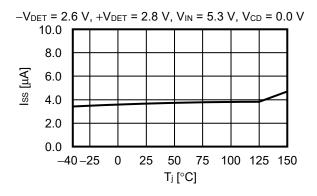


#### (2) $V_{OUT} = 5.0 V$

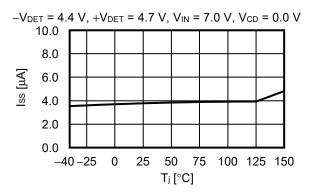


#### 3. 3. 2 S-19310 Series B type

#### (1) $V_{OUT} = 3.3 V$

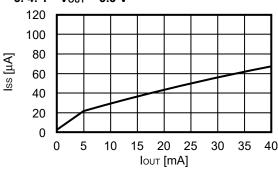


## (2) $V_{OUT} = 5.0 V$

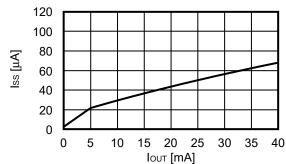


## 3. 4 Current consumption vs. Output current (Ta = +25°C)

3. 4. 1 V<sub>OUT</sub> = 3.3 V



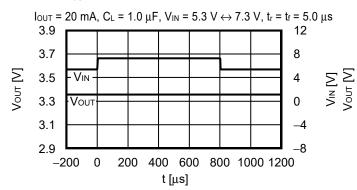
3. 4. 2 Vout = 5.0 V



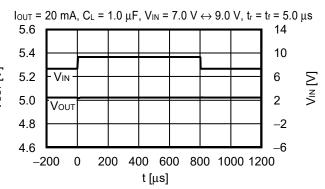
#### ■ Reference Data

#### 1. Transient response characteristics when input (Ta = +25°C)

#### 1. 1 Vout = 3.3 V

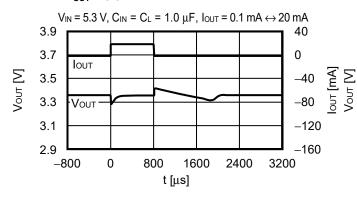


#### 1. 2 Vout = 5.0 V

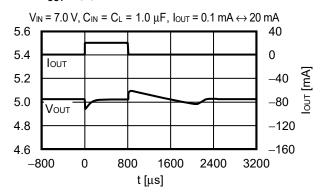


#### 2. Transient response characteristics of load (Ta = +25°C)

#### 2. 1 $V_{OUT} = 3.3 V$

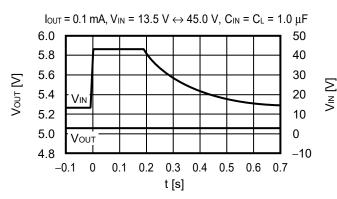


#### 2. 2 V<sub>OUT</sub> = 5.0 V



#### 3. Load dump characteristics (Ta = +25°C)

#### 3. 1 $V_{OUT} = 5.0 V$



4. Example of equivalent series resistance vs. Output current characteristics (Ta = +25°C)

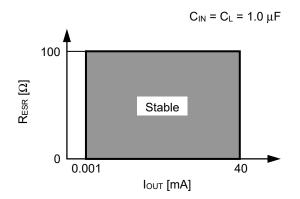
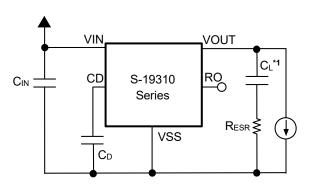


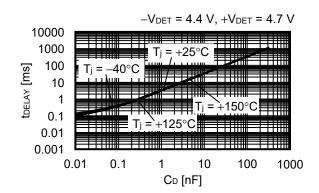
Figure 30



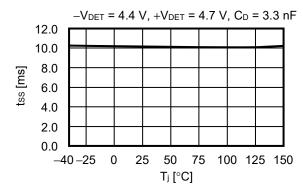
\*1.  $C_L$ : TDK Corporation CGA4J3X8R1C105K (1.0  $\mu F$ )

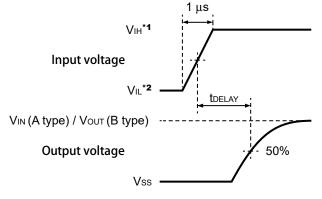
Figure 31

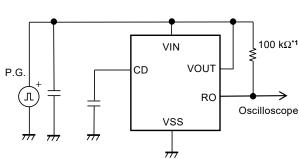
5. Release delay time vs. CD pin capacitance (Without output pin capacitance)



## 6. Release delay time vs. Junction temperature







\*1. V<sub>IH</sub> = V<sub>OUT(S)</sub>

- \*1. Only S-19310 Series A type
- \*2. When 2.6 V  $\leq$  +V<sub>DET(S)</sub> < 3.5 V, V<sub>IL</sub> = 2.5 V When 3.5 V  $\leq$  +V<sub>DET(S)</sub>  $\leq$  5.0 V, V<sub>IL</sub> = +V<sub>DET(S)</sub> 1.0 V

Figure 32 Test Condition of Release Delay Time

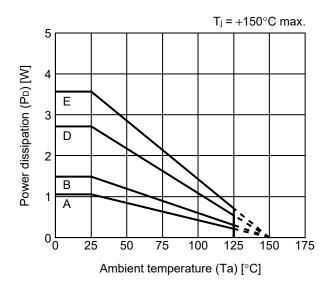
Figure 33 Test Circuit of Release Delay Time

Caution The above connection diagram and constants will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constants.

HTMSOP-8

## **■** Power Dissipation

## SOT-89-5



_		Ε												
Power dissipation (P <sub>D</sub> ) [W]	4	С												
ation	3	D												
dissip	2							$\downarrow$					_	
Power	1	В		_				$\stackrel{\cdot}{\downarrow}$	/					
_		Α			+			\$		14	•			
	0	)	2	5	50	7	5	100	0 12	25	15	0	17	5

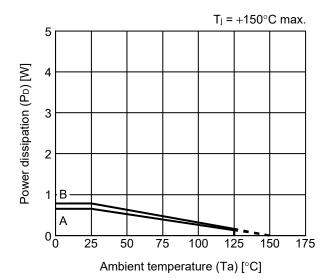
 $T_j = +150$ °C max.

Board	Power Dissipation (P <sub>D</sub> )
Α	1.05 W
В	1.49 W
С	_
D	2.72 W
Е	3.57 W

Board	Power Dissipation (P <sub>D</sub> )
Α	0.79 W
В	1.11 W
С	3.21 W
D	3.13 W
Е	4.17 W

Ambient temperature (Ta) [°C]

#### SOT-23-5



Board	Power Dissipation (P <sub>D</sub> )
Α	0.65 W
В	0.78 W
С	_
D	_
Е	_

# **SOT-89-5** Test Board

# (1) Board A



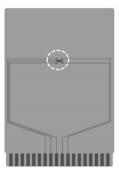
Thermal via

Item	S	pecification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		2
Copper foil layer [mm]	1	Land pattern and wiring for testing: t0.070
2		-
	3	-

74.2 x 74.2 x t0.070

( ) IC Mount Area

## (2) Board B



Item	S	pecification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
Copper foil layer [mm]	1	Land pattern and wiring for testing: t0.070
	2	74.2 x 74.2 x t0.035
3		74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		-

# (3) Board D



Item	S	pecification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
Copper foil layer [mm]	1	Pattern for heat radiation: 2000mm <sup>2</sup> t0.070
	2	74.2 x 74.2 x t0.035
	3	74.2 x 74.2 x t0.035
4		74.2 x 74.2 x t0.070
Thermal via		-

# (4) Board E



Item	S	pecification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
Copper foil layer [mm]	1	Pattern for heat radiation: 2000mm <sup>2</sup> t0.070
	2	74.2 x 74.2 x t0.035
	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		Number: 4 Diameter: 0.3 mm

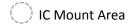


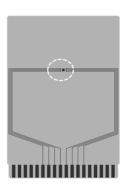
enlarged view

No. SOT895-A-Board-SD-1.0

# **HTMSOP-8** Test Board

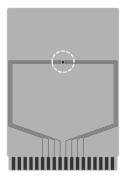
## (1) Board A





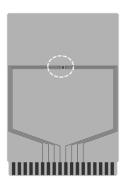
Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		2
	1	Land pattern and wiring for testing: t0.070
Coppor foil layer [mm]	2	-
Copper foil layer [mm]	3	-
	4	74.2 x 74.2 x t0.070
Thermal via		-

# (2) Board B



Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
Copper foil layer [mm]	1	Land pattern and wiring for testing: t0.070
	2	74.2 x 74.2 x t0.035
	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		-

# (3) Board C



Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
	1	Land pattern and wiring for testing: t0.070
Cappar fail layer [mm]	2	74.2 x 74.2 x t0.035
Copper foil layer [mm]	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		Number: 4 Diameter: 0.3 mm



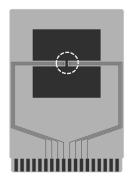
enlarged view

No. HTMSOP8-A-Board-SD-1.0

# **HTMSOP-8** Test Board

O IC Mount Area

## (4) Board D

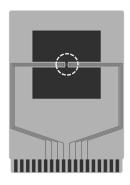


Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
Copper foil layer [mm]	1	Pattern for heat radiation: 2000mm <sup>2</sup> t0.070
	2	74.2 x 74.2 x t0.035
	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		-



enlarged view

## (5) Board E



Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
Copper foil layer [mm]	1	Pattern for heat radiation: 2000mm <sup>2</sup> t0.070
	2	74.2 x 74.2 x t0.035
	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		Number: 4 Diameter: 0.3 mm

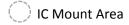


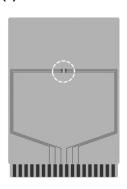
enlarged view

No. HTMSOP8-A-Board-SD-1.0

# **SOT-23-3/3S/5/6** Test Board

#### (1) Board A





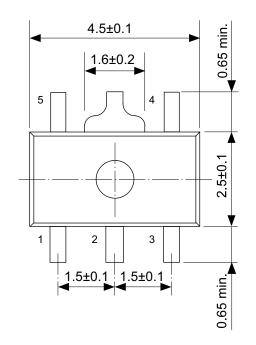
Item		Specification	
Size [mm]		114.3 x 76.2 x t1.6	
Material		FR-4	
Number of copper foil la	mber of copper foil layer 2		
Copper foil layer [mm]	1	Land pattern and wiring for testing: t0.070	
	2	-	
	3	-	
	4	74.2 x 74.2 x t0.070	
Thermal via		-	

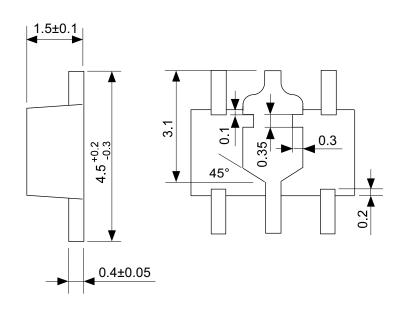
#### (2) Board B

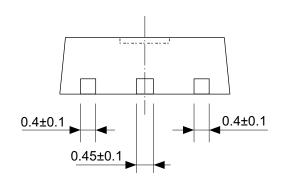


Item		Specification	
Size [mm]		114.3 x 76.2 x t1.6	
Material		FR-4	
Number of copper foil layer		4	
	1	Land pattern and wiring for testing: t0.070	
Copper foil layer [mm]	2	74.2 x 74.2 x t0.035	
Copper foil layer [min]	3	74.2 x 74.2 x t0.035	
	4	74.2 x 74.2 x t0.070	
Thermal via		-	

No. SOT23x-A-Board-SD-2.0

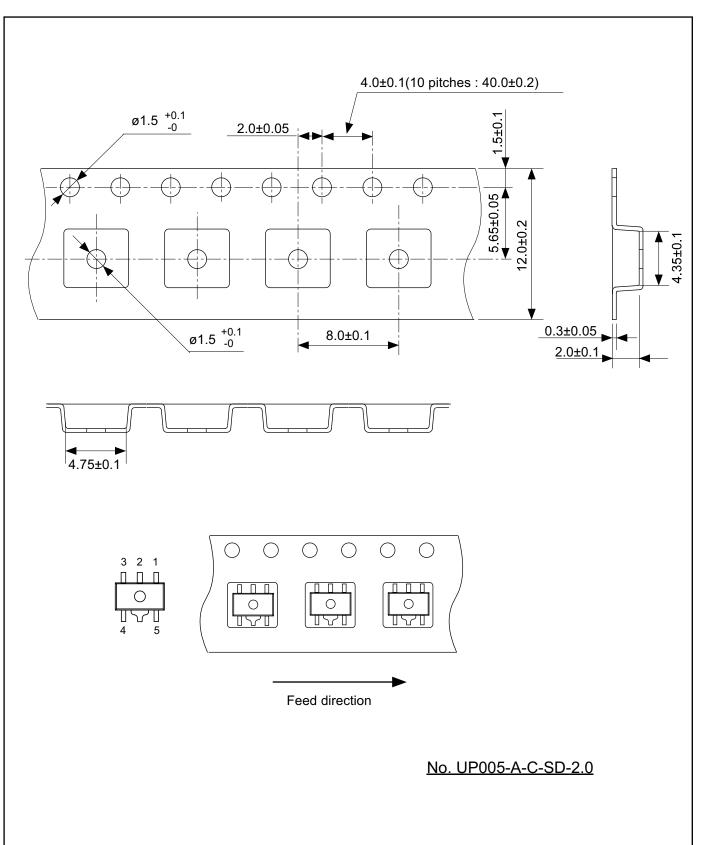




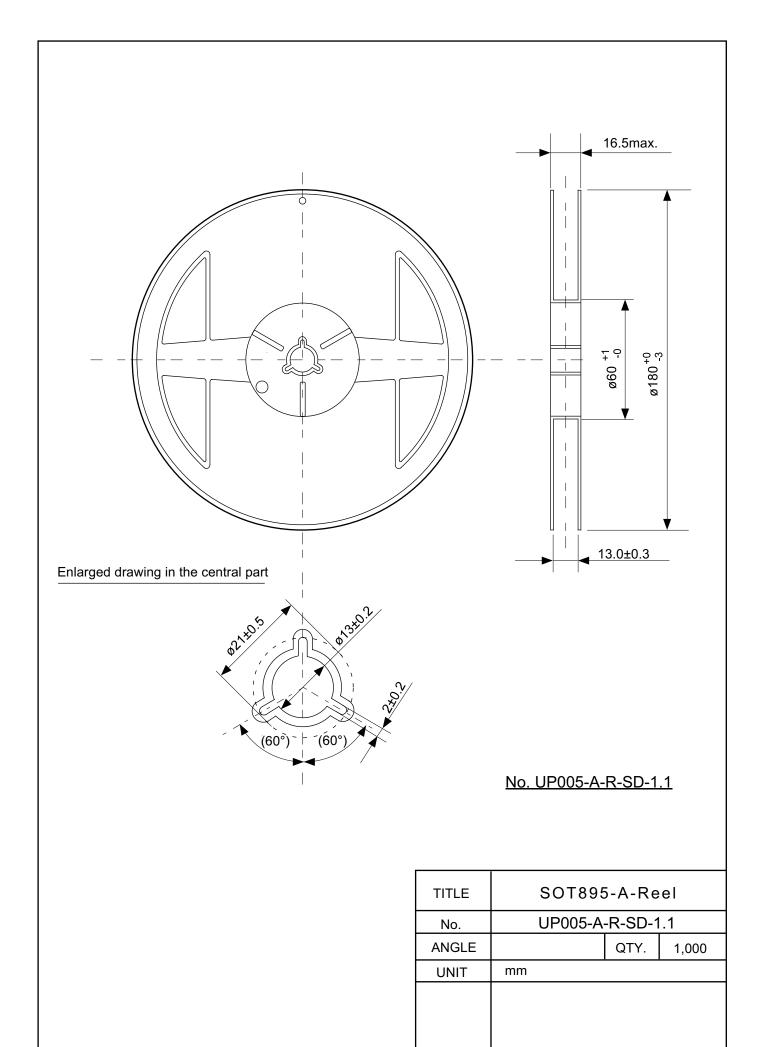


## No. UP005-A-P-SD-2.0

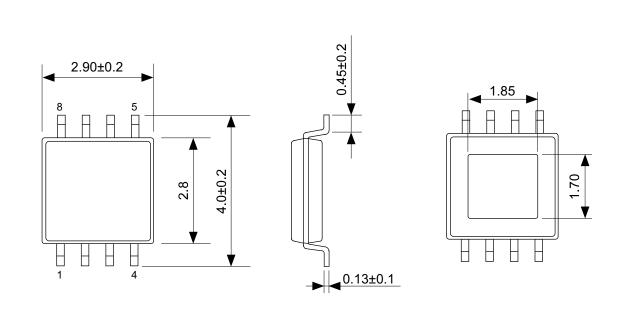
TITLE	SOT895-A-PKG Dimensions		
No.	UP005-A-P-SD-2.0		
ANGLE	$\Phi$		
UNIT	mm		
ABLIC Inc.			

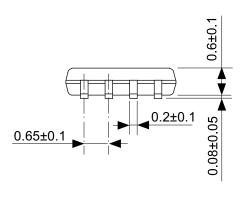


TITLE	SOT895-A-Carrier Tape	
No.	UP005-A-C-SD-2.0	
ANGLE		
UNIT	mm	
ABLIC Inc.		



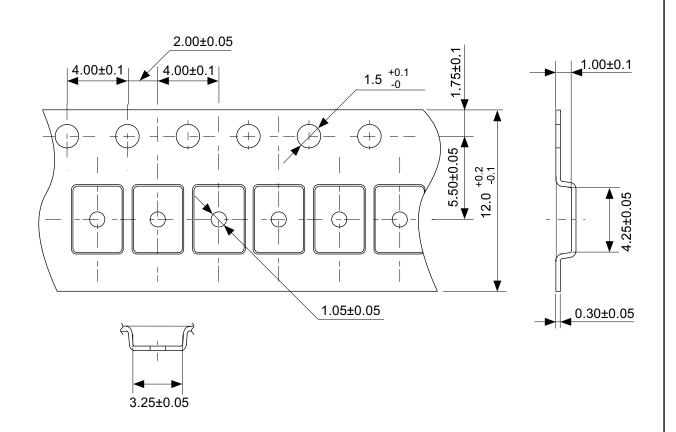
ABLIC Inc.

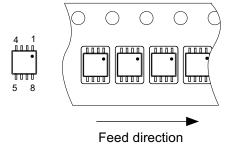




## No. FP008-A-P-SD-2.0

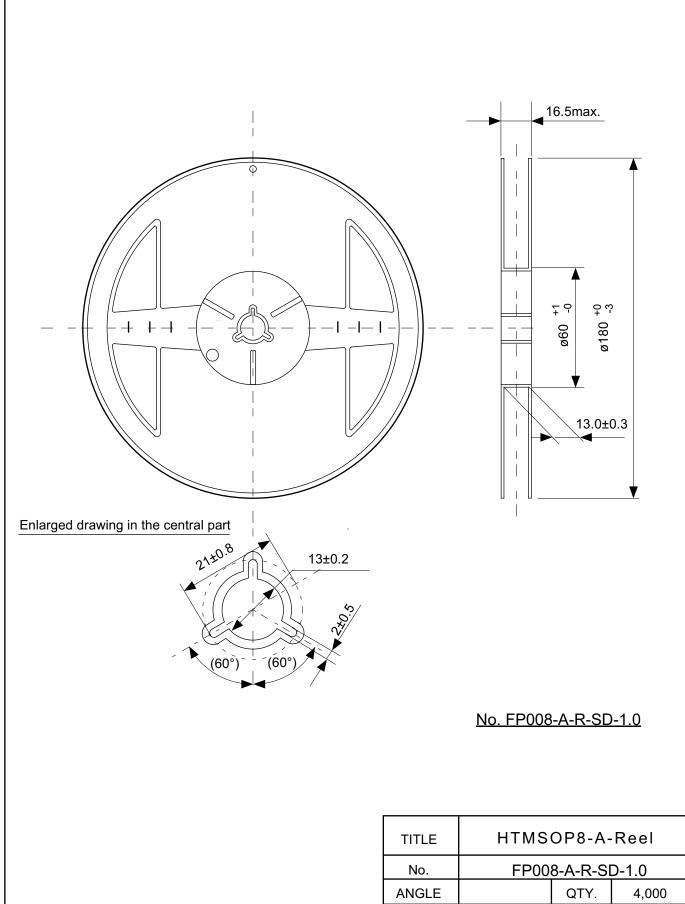
TITLE	HTMSOP8-A-PKG Dimensions	
No.	FP008-A-P-SD-2.0	
ANGLE	$\bigoplus$	
UNIT	mm	
ABLIC Inc.		



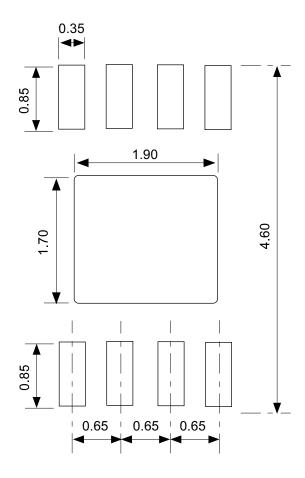


#### No. FP008-A-C-SD-1.0

TITLE	HTMSOP8-A-Carrier Tape		
No.	FP008-A-C-SD-1.0		
ANGLE			
UNIT	mm		
ABLIC Inc.			



TITLE	HTMSOP8-A-Reel		
No.	FP008-A-R-SD-1.0		
ANGLE	QTY. 4,000		
UNIT	mm		
ABLIC Inc.			



## No. FP008-A-L-SD-2.0

TITLE	HTMSOP8-A -Land Recommendation	
No.	FP008-A-L-SD-2.0	
ANGLE		
UNIT	mm	
ABLIC Inc.		







## No. MP005-A-P-SD-1.3

SOT235-A-PKG Dimensions			
MP005-A-P-SD-1.3			
<b>\$</b>			
mm			
ABLIC Inc.			



TITLE	SOT235-A-Carrier Tape	
No.	MP005-A-C-SD-2.1	
ANGLE		
UNIT	mm	
ABLIC Inc.		



TITLE	SOT235-A-Reel		
No.	MP005-A-R-SD-1.1		
ANGLE		QTY.	3,000
UNIT	mm		
ABLIC Inc.			

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